

## *EXAMPLES OF FEATURES FOR SOME TYPICAL VOWELS*

### **A11-A.1 INTRODUCTION**

Appendix 5 presents twelve vowels that illustrate the positions of the jaw, tongue, and lips; that is, the approximated vocal tract configuration in the form of a midsagittal outline (upper left of figure), along with the synthesized acoustic waveform (upper right), the vocal tract cross-sectional area function (lower left), and the vocal tract frequency response (lower right). These results were obtained using the articulatory speech synthesizer, which minimized the error between the vowel target formants and the articulatory model formants. The vowels and the error are listed here. Consult Appendix 5 for the other data.

- Vowel IY as in beet, error 0.01432%
- Vowel IH as in bit, 0.00632%
- Vowel EY as in bait, error 0.03832%
- Vowel EH as in bet, error 0.01204%
- Vowel AE as in bat, error 0.06795%
- Vowel UW as boot, error 0.02982%
- Vowel UH as in book, error 0.01867%
- Vowel OW as boat, error 0.19218%
- Vowel AO as in bought, error 0.05938%
- Vowel AA as in Bach, error 0.065087%
- Vowel AH as in but, error 0.21787%
- Vowel ER as in bird, error 0.57565%

# ACOUSTIC TRANSFER FUNCTION CALCULATION

The formant frequencies are required by the simulated annealing algorithm to calculate the articulatory-to-acoustic inverse transform. The formant frequencies need to be decomposed from the acoustic transfer function or from the input impedance, both of which are calculated from the vocal tract cross-sectional area function. This appendix presents a derivation of the acoustic transfer function for various structures using a circuit network representation and transmission matrix theory. In Appendix 11, Section A11.2.3.1.3, it was mentioned that an elemental tube with uniform area  $A$  and length  $l$  can be analogous to the transmission line circuit model. The circuit in Figure 5.1 and the definitions given in Figure A10.27 form the basis for the following derivations. The circuit in Figure 5.1 can be rearranged as a four terminal T network, as shown in Figure A11-B.1. This network representation facilitates the acoustic transfer function calculation, which is defined as the ratio of the total output volume velocity from the vocal tract and the nasal tract to the excitation volume velocity source input. Define the propagation constant as  $\gamma = \sqrt{zy}$ , where  $z = R + j\omega L$ ,  $y = G + j\omega C + \frac{1}{Z_w}$ , and  $Z_w = R_w + j\omega L_w + \frac{1}{j\omega C_w}$ . Let the characteristic impedance be

$$Z_a = Z_0 \tanh\left(\frac{\gamma \ell}{2}\right)$$

and

$$Z_b = \frac{Z_0}{\sinh(\lambda \ell)}$$

See Fant (1960) and Flanagan (1972a) for more details.

It is well known that the input-output characteristics of a four-terminal network are described by a matrix equation of the form

$$\begin{bmatrix} P_i(\omega) \\ U_i(\omega) \end{bmatrix} = [T(\omega)] \begin{bmatrix} P_o(\omega) \\ U_o(\omega) \end{bmatrix} \quad (\text{A11-B.1})$$

where  $P_i(\omega)$  and  $U_i(\omega)$  are the sound pressure and volume velocity, respectively, at the input of the network;  $P_o(\omega)$  and  $U_o(\omega)$  are the corresponding outputs of the network; and  $T(\omega)$  is the transmission matrix (also called ABCD matrix or chain matrix) of the four-terminal network. From Ohm's law and the current loop law, the transmission matrix of Figure A11-B.1 is

$$T(\omega) = \begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 + \frac{Z_a}{Z_b} & 2Z_a + \frac{Z_a^2}{Z_b} \\ \frac{1}{Z_b} & 1 + \frac{Z_a}{Z_b} \end{bmatrix} \quad (\text{A11-B.2})$$

Now, consider two elemental sections with a shunt impedance  $Z_s$  in between and with radiation impedance  $Z_r$  as load. We represent the radiation impedance as  $Z_r = R_r + j\omega L_r$ , where  $R_r = \frac{128gc}{9\pi^2 A_m}$ ,  $L_r = \frac{8g}{3\pi\sqrt{\pi} A_m}$ , and  $A_m$  is the lip or nostril opening area. The network representation is constructed in Figure A11-B.2. Note that the shunt impedance  $Z_s$  can be considered as the modelling circuit for the nasal sinus (see Appendix A11). Assume that the transmission matrices of section  $i-1$  and section  $i$

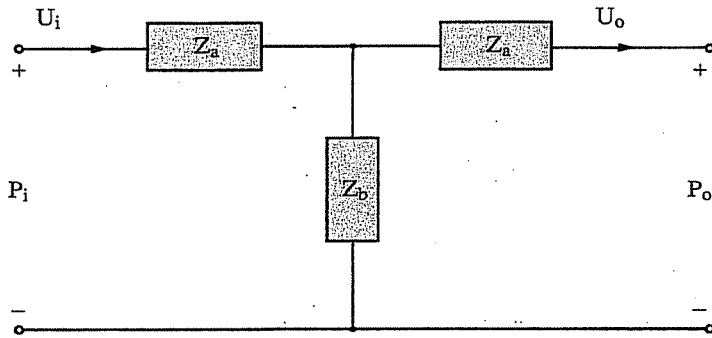


FIGURE A11-B.1 Four-terminal T-network for a uniform elemental tube.

are  $T_{i-1}(\omega)$  and  $T_i(\omega)$  respectively. Forming a dummy four-terminal T-network as in Figure A11-B.1 with  $Z_a = 0$  and  $Z_b = Z_s$  for the shunt element  $Z_s$  and applying Equation (A11-B.2), the transmission matrix is

$$T_s(\omega) = \begin{bmatrix} A_s & B_s \\ C_s & D_s \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \frac{1}{Z_s} & 1 \end{bmatrix} \tag{A11-B.3}$$

From the theory of cascade circuit networks, the overall transmission matrix for Figure A11-B.2 is the product of the individual transmission matrices for the sections. Thus the relation between pressure and volume velocity at the input  $P_{i-1}$ ,  $U_{i-1}$  and at the radiation  $P_r$ ,  $U_r$  can be written as

$$\begin{bmatrix} P_{i-1} \\ U_{i-1} \end{bmatrix} = T_{i-1}(\omega) \cdot T_s(\omega) \cdot T_i(\omega) \begin{bmatrix} P_r \\ U_r \end{bmatrix} = \begin{bmatrix} A_f & B_f \\ C_f & D_f \end{bmatrix} \begin{bmatrix} P_r \\ U_r \end{bmatrix} \tag{A11-B.4}$$

From this matrix equation and Ohm's law,  $P_r = Z_r U_r$ , the transfer function  $H(\omega)$  from  $U_{i-1}$  to  $U_r$  is

$$H(\omega) = \frac{U_r(\omega)}{U_{i-1}(\omega)} = \frac{1}{C_f Z_r + D_f} \tag{A11-B.5}$$

and the input impedance  $Z_{in}$  is

$$Z_{in}(\omega) = \frac{P_{i-1}(\omega)}{U_{i-1}(\omega)} = \frac{A_f Z_r + B_f}{C_f Z_r + D_f} \tag{A11-B.6}$$

By use of the four-terminal network representation and the transmission matrix, we can calculate the acoustic transfer function of the vocal tract system for human speech production. Since the excitation source can be at the glottis or inside the vocal tract, there are four cases.

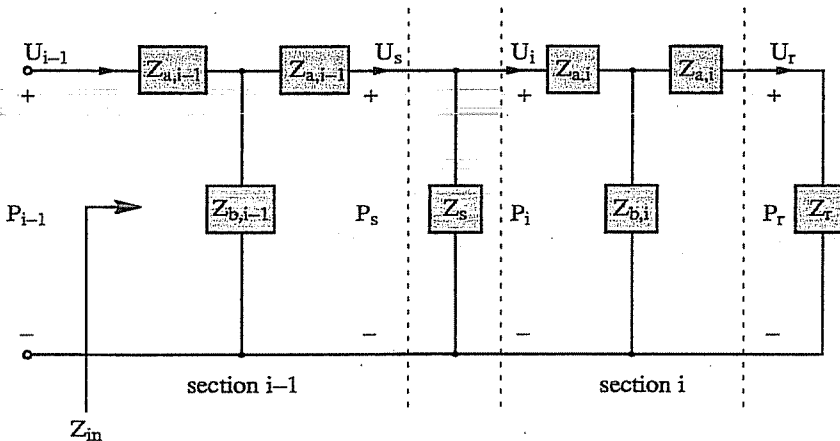


FIGURE A11-B.2 Network representation of a two-section tube with a shunt element in between and a radiation load.

**Case A: Excitation Source Located at the Glottis.** Let  $Z_{or}$  and  $Z_{nr}$  be the radiation impedances of the vocal tract and the nasal tract, respectively. Let  $Z_{sub}$  as the impedance of the glottis and the subglottal system looking backward from the glottis. Use  $A_p B_p C_p D_p$ ,  $A_o B_o C_o D_o$ , and  $A_n B_n C_n D_n$  to represent the transmission matrices of the pharyngeal tube, oral tract, and nasal tract, respectively. Figure A11-B.3(a) shows the equivalent network representation of the vocal system.

To calculate the acoustic transfer function  $H(\omega)$  from  $U_g$  to  $U_r = U_{or} + U_{nr}$ , we first compute the input impedance of the oral tract and the input impedance of the nasal tract, both of which are seen downstream from the bifurcation point. From Equation (A11-B.6), we have the input impedance of oral tract as

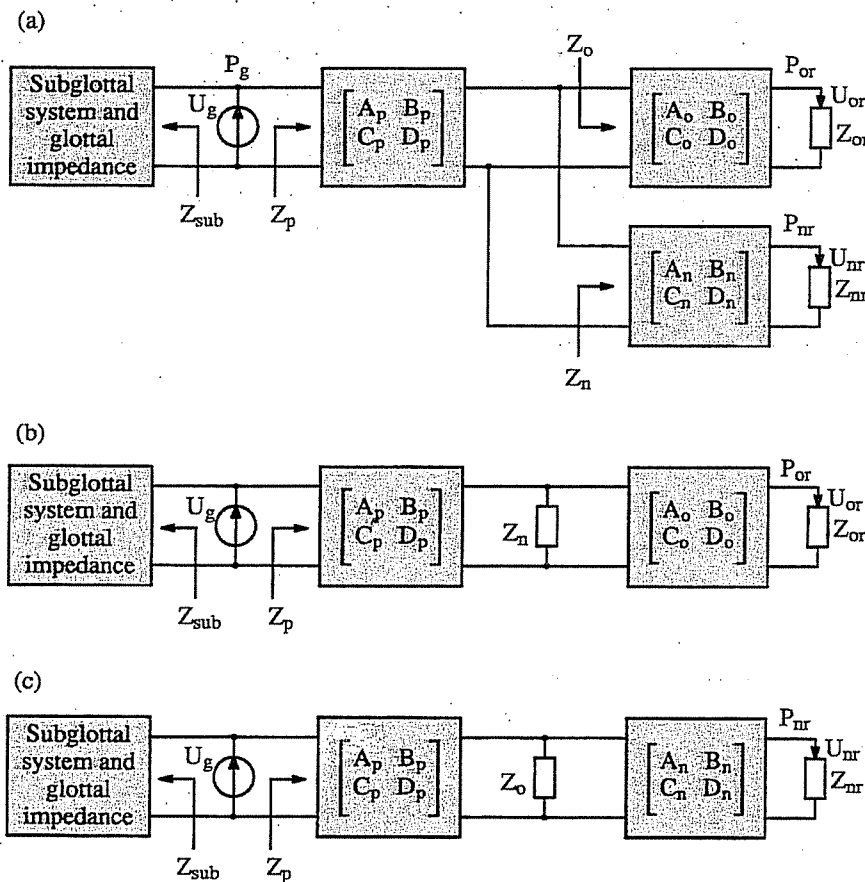
$$Z_o = \frac{A_o Z_{or} + B_o}{C_o Z_{or} + D_o} \tag{A11-B.7}$$

and the nasal tract input impedance as

$$Z_n = \frac{A_n Z_{nr} + B_n}{C_n Z_{nr} + D_n} \tag{A11-B.8}$$

The second step is to calculate the acoustic transfer function  $H_v(\omega)$  of the vocal tract from  $U_g$  to  $U_{or}$ . Forming an equivalent network for Figure A11-B.3(a) (see Figure A11-B.3(b)) and applying the analysis results of Figure A11-B.2 (refer to Equation A11-B.5), we have

$$H_v = \frac{Z_{sub}}{Z_p + Z_{sub}} \cdot \frac{1}{C_{vv} Z_{or} + D_{vv}} \tag{A11-B.9}$$



**FIGURE A11-B.3** Case A with excitation source located at the glottis. (a) Network representation of vocal system. (b) Network representation for calculating  $H_v(\omega)$ . (c) Network representation for calculating  $H_n(\omega)$ .

and the corresponding transmission matrix as

$$\begin{bmatrix} A_{vv} & B_{vv} \\ C_{vv} & D_{vv} \end{bmatrix} = \begin{bmatrix} A_p & B_p \\ C_p & D_p \end{bmatrix} \begin{bmatrix} 1 & 0 \\ \frac{1}{Z_n} & 1 \end{bmatrix} \begin{bmatrix} A_o & B_o \\ C_o & D_o \end{bmatrix} \quad (\text{A11-B.10})$$

Similarly, from Equation (A11-B.7), the input impedance of the vocal tract seen downstream from the glottis is

$$Z_p = \frac{A_{vv}Z_{or} + B_{vv}}{C_{vv}Z_{or} + D_{vv}} \quad (\text{A11-B.11})$$

Applying the same procedure to the Figure A11-B.3(c), the acoustic transfer function  $H_n(\omega)$  from  $U_g$  to the output of the nasal tract  $U_{nr}$  is

$$H_n = \frac{Z_{sub}}{Z_p + Z_{sub}} \cdot \frac{1}{C_{nn}Z_{nr} + D_{nn}} \quad (\text{A11-B.12})$$

and the corresponding transmission matrix is

$$\begin{bmatrix} A_{nn} & B_{nn} \\ C_{nn} & D_{nn} \end{bmatrix} = \begin{bmatrix} A_p & B_p \\ C_p & D_p \end{bmatrix} \begin{bmatrix} 1 & 0 \\ \frac{1}{Z_o} & 1 \end{bmatrix} \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} \quad (\text{A11-B.13})$$

The vocal system acoustic transfer function is  $H(\omega) = H_v(\omega) + H_n(\omega)$ .

**Case B: Excitation Source Located in the Pharyngeal Tube.** Let  $A_{pl}B_{pl}C_{pl}D_{pl}$  represent the transmission matrix of the lower part of the pharynx with the input at the excitation location and output is at the glottis, as shown in Figure A11-B.4. The results obtained in Case A can be applied to this case. Replace the  $Z_{sub}$  in Equation (A11-B.9) and (A11-B.12) with the  $Z'_{sub}$ , where the  $Z'_{sub}$  is calculated as follows.

$$Z'_{sub} = \frac{A_{pl}Z_{sub} + B_{pl}}{C_{pl}Z_{sub} + D_{pl}} \quad (\text{A11-B.14})$$

Then, the sum of Equations (A11-B.9) and (A11-B.12) yields the final acoustic transfer function.

**Case C: Excitation Source Located at the Bifurcation of the Vocal Tract and the Nasal Tract.** As shown in Figure A11-B.5, the network representation is the same as Figure A11-B.3 except the excitation source is located at the bifurcation of the vocal tract and the nasal tract. First, the input impedances  $Z_o$  and  $Z_n$  are calculated as in Equations (A11-B.7) and (A11-B.8). Then  $H_v(\omega)$  and  $H_n(\omega)$  are obtained as follows.

$$H_v = \frac{Z'_{sub}Z_n}{Z'_{sub}Z_o + Z_oZ_n + Z_nZ'_{sub}} \cdot \frac{1}{C_oZ_{or} + D_o} \quad (\text{A11-B.15})$$

$$H_n = \frac{Z'_{sub}Z_o}{Z'_{sub}Z_o + Z_oZ_n + Z_nZ'_{sub}} \cdot \frac{1}{C_nZ_{nr} + D_n} \quad (\text{A11-B.16})$$

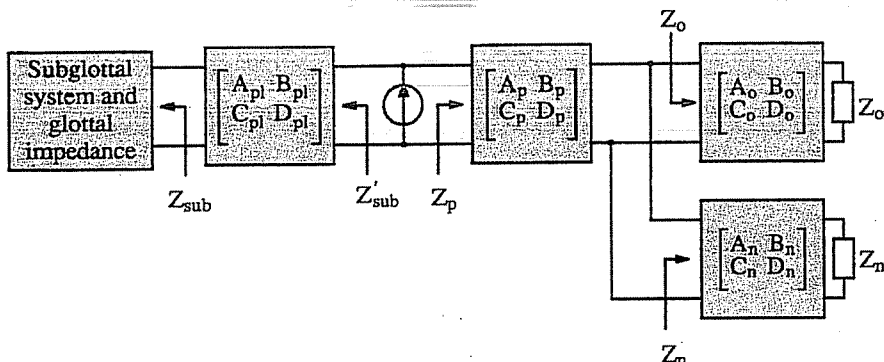


FIGURE A11-B.4 Case B with the excitation source located in the pharyngeal tube.

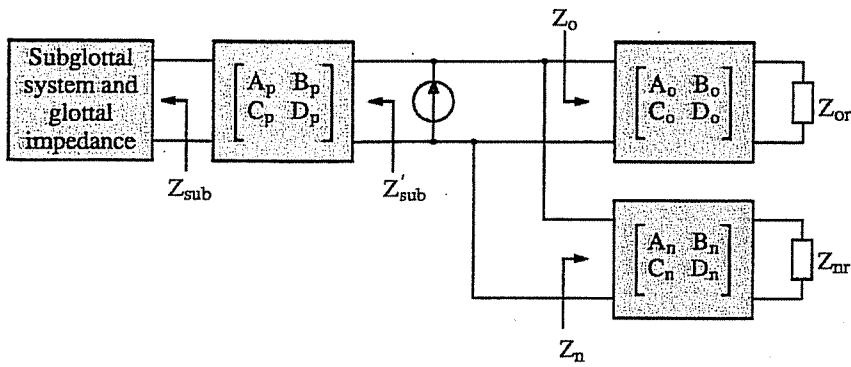


FIGURE A11-B.5 Case C with the excitation source located at the bifurcation point.

where the impedance  $Z'_{sub}$  is

$$Z'_{sub} = \frac{A_p Z_{sub} + B_p}{C_p Z_{sub} + D_p} \tag{A11-B.17}$$

The sum of  $H_v(\omega)$  and  $H_n(\omega)$  yields the acoustic transfer function for the vocal system.

**Case D: Excitation Source Located in the Oral Tract.** As shown in Figure A11-B.6(a), we use  $A_{ob}B_{ob}C_{ob}D_{ob}$  and  $A_{of}B_{of}C_{of}D_{of}$  to represent the transmission matrix of the back and front parts of oral tract, respectively. From Equations (A11-B.7), (A11-B.8), and (A11-B.17), we

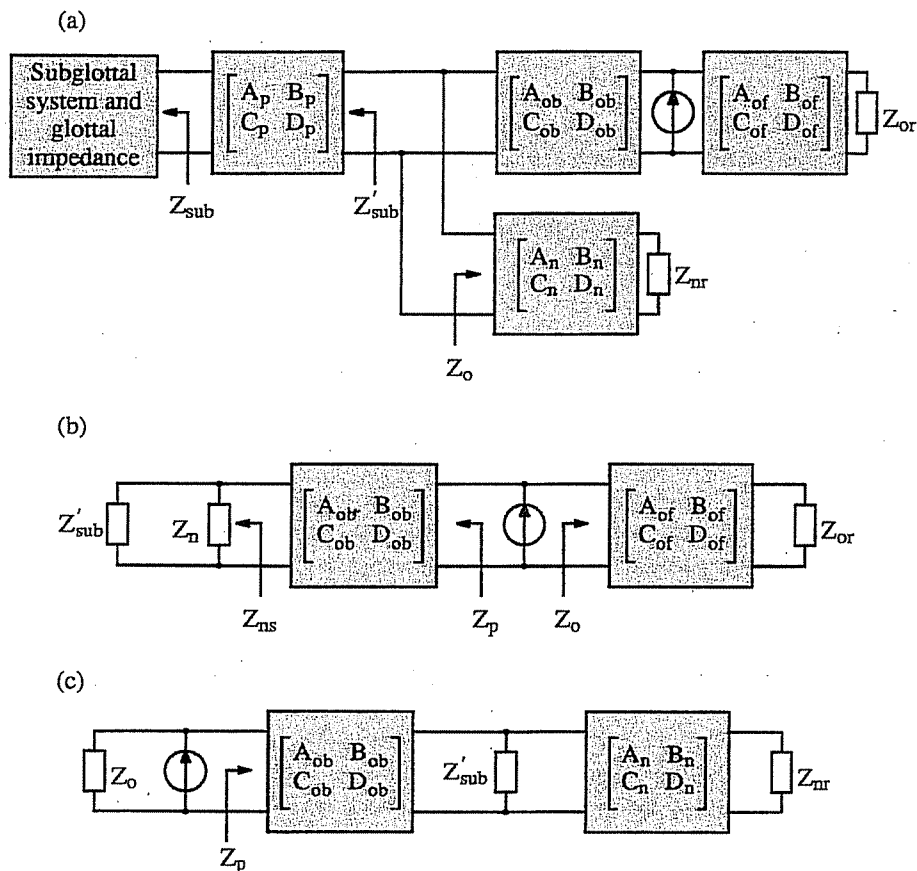


FIGURE A11-B.6 Case D with excitation source located in the oral tract. (a) Network representation of vocal system. (b) Network representation for calculating  $H_v(\omega)$ . (c) Network representation for calculating  $H_n(\omega)$ .

calculate the impedances  $Z_o$ ,  $Z_n$ , and  $Z'_{sub}$  as follows.

$$Z_o = \frac{A_{of}Z_{or} + B_{of}}{C_{of}Z_{or} + D_{of}} \tag{A11-B.18}$$

$$Z_n = \frac{A_nZ_{nr} + B_n}{C_nZ_{nr} + D_n} \tag{A11-B.19}$$

$$Z'_{sub} = \frac{A_pZ_{sub} + B_p}{C_pZ_{sub} + D_p} \tag{A11-B.20}$$

If we define  $Z_{ns}$  (see Figure A11-B.6(b)) as

$$Z_{ns} = \frac{Z_nZ'_{sub}}{Z_n + Z'_{sub}} \tag{A11-B.21}$$

then the impedance  $Z_p$  can be calculated as

$$Z_p = \frac{A_{ob}Z_{ns} + B_{ob}}{C_{ob}Z_{ns} + D_{ob}} \tag{A11-B.22}$$

Now, apply Equation (A11-B.9) to obtain  $H_v(\omega)$  as follows.

$$H_v = \frac{Z_p}{Z_p + Z_o} \cdot \frac{1}{C_{of}Z_{or} + D_{of}} \tag{A11-B.23}$$

For  $H_n(\omega)$ , we form the network as shown in Figure A11-B.6(c) and apply Equation (A11-B.12) to yield

$$H_n = \frac{Z_o}{Z_p + Z_o} \cdot \frac{1}{C_{on}Z_{nr} + D_{on}} \tag{A11-B.24}$$

where the corresponding transmission matrix is

$$\begin{bmatrix} A_{on} & B_{on} \\ C_{on} & D_{on} \end{bmatrix} = \begin{bmatrix} A_{ob} & B_{ob} \\ C_{ob} & D_{ob} \end{bmatrix} \begin{bmatrix} 1 & 0 \\ \frac{1}{Z'_{sub}} & 1 \end{bmatrix} \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} \tag{A11-B.25}$$

Then, the final acoustic transfer function is  $H_v(\omega) + H_n(\omega)$ .

We have considered the acoustic transfer function of the soft-wall, lossy vocal system. For some cases, calculation of the acoustic transfer function of a lossless or lossy, hard-wall vocal system is adequate and also easier. For a rigid and lossy vocal system, the wall impedance  $Z_w$  of each section is removed, causing the impedance  $Z_w$  to be infinity. In this case, the admittance  $y$  becomes  $y = G + j\omega C$ . For a rigid and lossless vocal system, this means that no dissipation elements and no wall impedance are present for each section. Then the propagation constant  $\gamma$  becomes an imaginary quantity.

In the above derivations, we assume that impedance  $Z_{sub}$  is composed of the Foster-chain circuit model for the subglottal system with glottal impedance  $Z_g = R_g + j\omega L_g$ , where  $R_g$  and  $L_g$  are defined in Appendix A11. Since the glottal area is a time-varying function, we assume that the glottal area and glottal volume velocity are constants when we evaluate the acoustic transfer function. In Case A, if the subglottal system and the glottal impedance are not included, then we let the impedance  $Z_{sub} \rightarrow \infty$ , which corresponds to an open circuit, that is, the block representing the subglottal system and the glottal impedance in Figure A11-B.3 are disconnected. For other cases, two glottal conditions are considered. If the glottis is open, the effects of the subglottal system and the glottal impedance are evaluated. Otherwise, if the glottis is closed, we let  $Z_{sub} = 0$ , that is,  $Z_{sub}$  is a short circuit. The velopharyngeal port opening area, which couples the nasal tract to the vocal system, also affects the acoustic transfer function. Our software system allows the user to change the velopharyngeal port opening area, the number of coupled sinus cavities, and the location of the excitation source when evaluating the acoustic transfer function.

# DERIVATION OF DISCRETE TIME ACOUSTIC EQUATIONS

As covered in Section A11.2.3.1, the vocal tract tube can be described by two coupled partial differential acoustic equations. These two acoustic equations are functions of both time and space. Approximating the vocal tract as a sequence of elemental sections corresponds to digitizing the vocal tract in space, that is, spatial sampling. For each elemental section, the transmission-line analog approach is applied to form the equivalent circuit model, as seen in Figure 5.1. Connecting the equivalent circuit of each section together in combination with the equivalent circuit models of the other parts of the vocal system (subglottal system, glottis, and nasal sinus cavities), a lumped circuit network representation of the vocal system can be formed, as shown in Figure A11.18; the time-domain approach, the Kirchoff's and Ohm's laws are applied to the circuit network to obtain sets of differential equations. These differential equations, which correspond to the equivalent acoustic equations, which govern the generation and the propagation of acoustic waves inside the vocal system, are transformed into discrete-time representations. This appendix provides a detailed derivation of the discrete-time acoustic equations, that is, the difference matrix equations. The discretization scheme is similar to the work of Maeda (1982a). Our model, however, provides more features, such as the subglottal system, nasal sinus cavities, and turbulence noise source.

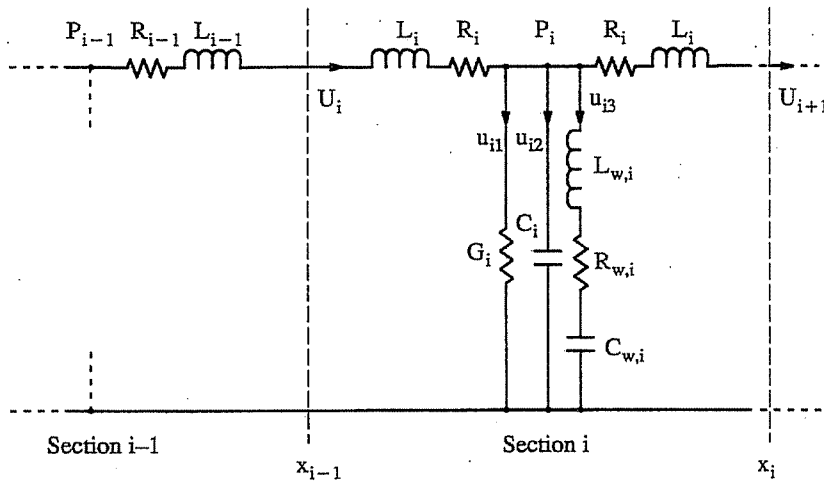
Consider the transmission-line circuit model, shown in Figure A11-C.1, of the  $i$ th section of the vocal tract. Define the volume velocity (current) at the input of the  $i$ th section, that is, at  $x_{i-1}$  in space, as  $U_i(t) = u_i(x_{i-1}, t)$ , where  $x_{i-1} = \sum_{k=1}^{i-1} l_k$  is the vocal tract length from the glottis to the  $i$ th section. Similarly, define the central pressure (voltage) of the  $i$ th section as  $P_i(t) = p_i(x_i - \frac{l_i}{2}, t)$ , where  $l_i$  is the section length of section  $i$ . From Kirchoff's and Ohm's laws, we have the following differential equations.

$$\begin{aligned}
 P_{i-1}(t) - P_i(t) &= \frac{d}{dt} \{ [L_{i-1}(t) + L_i(t)] U_i(t) \} + [R_{i-1}(t) + R_i(t)] U_i(t) \\
 U_i(t) - U_{i+1}(t) &= u_{i1}(t) + u_{i2}(t) + u_{i3}(t) \\
 &= G_i(t) P_i(t) + \frac{d}{dt} \{ C_i(t) P_i(t) \} + u_{i3}(t) \\
 P_i(t) &= R_{w,i}(t) u_{i3}(t) + \frac{d}{dt} \{ L_{w,i}(t) u_{i3}(t) \} + \int_0^t \left\{ \frac{u_{i3}(\tau)}{C_{w,i}(\tau)} \right\} d\tau
 \end{aligned}
 \tag{A11-C.1}$$

There are three terms in Equation (A11-C.1); simple, differential, and integral terms. Define the general forms of these three terms as follows.

$$\begin{aligned}
 y_1(t) &\equiv c_1(t)x(t) \\
 y_2(t) &\equiv \frac{d}{dt} \{ c_2(t)x(t) \} \\
 y_3(t) &\equiv \int_0^t \{ c_3(\tau)x(\tau) \} d\tau
 \end{aligned}
 \tag{A11-C.2}$$





**FIGURE A11-C.1** A lumped transmission-line circuit model of the  $i$ th section.

where  $c_i(t)$  ( $i = 1, 2, \text{ or } 3$ ) is a coefficient that represents the time-varying circuit component, and  $x(t)$  can be  $P_i(t)$ ,  $U_i(t)$ , or  $u_{i3}(t)$ . Let  $y_i(n) = y_i(t = nT)$ ,  $c_i(n) = c_i(t = nT)$ , and  $x(n) = x(t = nT)$  represent the sampled values of  $y_i(t)$ ,  $c_i(t)$ , and  $x(t)$ , respectively, at  $t = nT$ , where  $n = 0, 1, 2, \dots$ , and  $T$  denotes the sampling time interval. For the first term, it is obvious that

$$y_1(n) = c_1(n)x(n) \quad (\text{A11-C.3})$$

For the differential term, we first integrate both sides

$$\int_{(n-1)T}^{nT} y_2(t) dt = \int_{(n-1)T}^{nT} \frac{d}{dt} \{c_2(t)x(t)\} dt = \int_{(n-1)T}^{nT} d\{c_2(t)x(t)\} \quad (\text{A11-C.4})$$

The trapezoidal rule is applied to the left hand side of Equation (A11-C.4) to yield

$$\int_{(n-1)T}^{nT} y_2(t) dt \equiv \frac{T}{2} [y_2(n) + y_2(n-1)] \quad (\text{A11-C.5})$$

Then the discrete-time form of Equation (A11-C.4) is

$$\frac{T}{2} [y_2(n) + y_2(n-1)] = c_2(n)x(n) - c_2(n-1)x(n-1) \quad (\text{A11-C.6})$$

Equation (A11-C.6) can be rewritten in recursive form as

$$y_2(n) = \frac{2}{T} c_2(n)x(n) - Q(n-1) \quad (\text{A11-C.7})$$

where

$$Q(n-1) = \frac{4}{T} c_2(n-1)x(n-1) - Q(n-2)$$

Similarly, the integral term is approximated by

$$y_3(n) = \frac{T}{2} c_3(n)x(n) + V(n-1) \quad (\text{A11-C.8})$$

where

$$V(n-1) = Tc_3(n-1)x(n-1) + V(n-2)$$

Now applying the rules of Equations (A11-C.3), (A11-C.7), and (A11-C.8) to the set of differential Equations (A11-C.1), the transformed equations are

$$\begin{aligned} P_{i-1}(n) - P_i(n) &= \left\{ R_{i-1}(n) + R_i(n) + \frac{2}{T} [L_{i-1}(n) + L_i(n)] \right\} U_i(n) - Q_{L_{i-1}L_i}(n-1) \\ U_i(n) - U_{i+1}(n) &= \left\{ G_i(n) + \frac{2}{T} C_i(n) \right\} P_i(n) - Q_{C_i}(n-1) + u_{i3}(n) \\ P_i(n) &= \left\{ R_{w,i}(n) + \frac{2}{T} L_{w,i}(n) + \frac{T}{2} \frac{1}{C_{w,i}(n)} \right\} u_{i3}(n) + V_{C_{w,i}}(n-1) - Q_{L_{w,i}}(n-1) \end{aligned} \quad (\text{A11-C.9})$$

where

$$\begin{aligned}
 Q_{L_{i-1}L_i}(n-1) &= \frac{4}{T}[L_{i-1}(n-1) + L_i(n-1)]U_i(n-1) - Q_{L_{i-1}L_i}(n-2) \\
 Q_{C_i}(n-1) &= \frac{4}{T}C_i(n-1)P_i(n-1) - Q_{C_i}(n-2) \\
 V_{C_{w,i}}(n-1) &= \frac{T}{C_{w,i}(n-1)}u_{i3}(n-1) + V_{C_{w,i}}(n-2) \\
 Q_{L_{w,i}}(n-1) &= \frac{4}{T}L_{w,i}(n-1)u_{i3}(n-1) - Q_{L_{w,i}}(n-2)
 \end{aligned}$$

Define

$$\begin{aligned}
 a_i(n) &= \frac{2}{T}L_i(n) + R_i(n) \\
 Y_{w,i}(n) &= \frac{1}{\frac{2}{T}L_{w,i}(n) + R_{w,i}(n) + \frac{T}{2} \frac{1}{C_{w,i}(n)}} \\
 b_i(n) &= \frac{1}{\frac{2}{T}C_i(n) + G_i(n) + Y_{w,i}(n)} \\
 V_i(n-1) &= Q_{C_i}(n-1) - Y_{w,i}(n)[Q_{L_{w,i}}(n-1) - V_{C_{w,i}}(n-1)] \\
 u_{i3}(n) &= Y_{w,i}(n)[P_i(n) + Q_{L_{w,i}}(n-1) - V_{C_{w,i}}(n-1)] \\
 H_i(n) &= a_{i-1}(n) + a_i(n) + b_{i-1}(n) + b_i(n) \\
 F_i(n) &\equiv b_{i-1}(n)V_{i-1}(n-1) + Q_{L_{i-1}L_i}(n-1) - b_i(n)V_i(n-1)
 \end{aligned} \tag{A11-C.10}$$

After some manipulations, we obtain the difference equations for the *i*th section, as follows.

$$\begin{aligned}
 F_i(n) &= -b_{i-1}(n)U_{i-1}(n) + H_i(n)U_i(n) - b_i(n)U_{i+1}(n) \\
 P_i(n) &= b_i(n)[U_i(n) - U_{i+1}(n) + V_i(n-1)]
 \end{aligned} \tag{A11-C.11}$$

Next, consider a shunt element, which models the nasal sinus, inserted in between two sections, *i* and *i* + 1. Figure A11-C.2 is the corresponding circuit model. The set of differential equations is

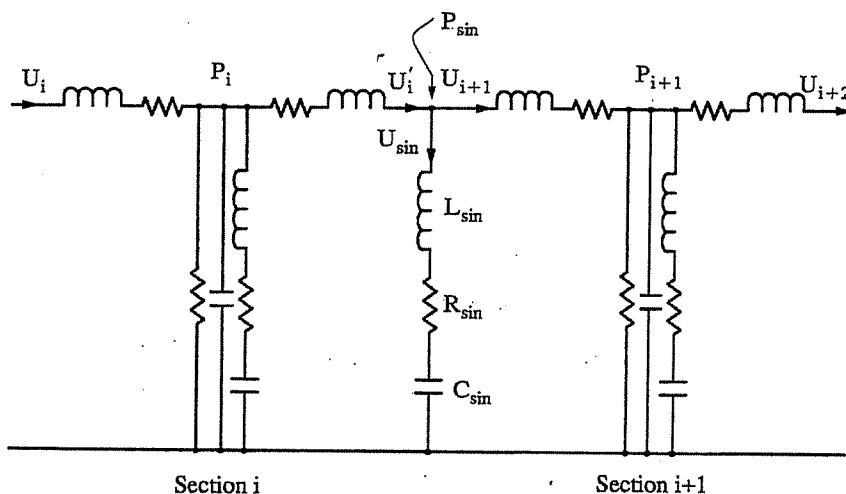


FIGURE A11-C.2 A shunt element of the nasal sinus inserted between two sections.

given by

$$\begin{aligned}
 P_i(t) - P_{\sin}(t) &= \frac{d}{dt} [L_i(t)U'_i(t)] + R_i(t)U'_i(t) \\
 P_{\sin}(t) - P_{i+1}(t) &= \frac{d}{dt} [L_{i+1}(t)U_{i+1}] + R_{i+1}(t)U_{i+1}(t) \\
 P_{\sin}(t) &= R_{\sin}U_{\sin}(t) + L_{\sin} \frac{dU_{\sin}(t)}{dt} + \frac{1}{C_{\sin}} \int_0^t U_{\sin}(\tau) d\tau
 \end{aligned}
 \tag{A11-C.12}$$

Applying the rules of Equations (A11-C.3), (A11-C.7), and (A11-C.8) to Equation set (A11-C.12), the corresponding difference equations are

$$\begin{aligned}
 P_i(n) - P_{\sin}(n) &= a_i(n)U'_i(n) - Q'_{L_i}(n-1) \\
 P_{\sin}(n) - P_{i+1}(n) &= a_{i+1}(n)U_{i+1}(n) - Q_{L_{i+1}}(n-1) \\
 P_{\sin}(n) &= b_{\sin} [U'_i(n) - U_{i+1}(n)] - Q_{L_{\sin}}(n-1) + V_{C_{\sin}}(n-1)
 \end{aligned}
 \tag{A11-C.13}$$

where

$$\begin{aligned}
 a_i(n) &= \frac{2}{T}L_i(n) + R_i(n) \\
 U_{\sin}(n) &= U'_i(n) - U_{i+1}(n) \\
 Q'_{L_i}(n-1) &= \frac{4}{T}L_i(n-1)U'_i(n-1) - Q'_{L_i}(n-2) \\
 Q_{L_{i+1}}(n-1) &= \frac{4}{T}L_{i+1}(n-1)U_{i+1}(n-1) - Q_{L_{i+1}}(n-2) \\
 Q_{L_{\sin}}(n-1) &= \frac{4}{T}L_{\sin}U_{\sin}(n-1) - Q_{L_{\sin}}(n-2) \\
 V_{C_{\sin}}(n-1) &= \frac{4}{C_{\sin}}U_{\sin}(n-1) + V_{C_{\sin}}(n-2) \\
 b_{\sin} &= \left[ \frac{2}{T}L_{\sin} + R_{\sin} + \frac{T}{2C_{\sin}} \right]
 \end{aligned}$$

After rearranging items, the difference equations governing the shunt element can be rewritten as

$$\begin{aligned}
 F_{\sin}(n) &= -b_i(n)U_i(n) + H_{sf}(n)U'_i(n) - b_{\sin}U_{i+1}(n) \\
 F_{i+1}(n) &= -b_{\sin}U'_i(n) + H_{sb}(n)U_{i+1}(n) - b_{i+1}(n)U_{i+2}(n) \\
 P_{\sin}(n) &= b_{\sin} [U'_i(n) - U_{i+1}(n)] - Q_{L_{\sin}}(n-1) + V_{C_{\sin}}(n-1) \\
 P_i(n) &= b_i(n) [U_i(n) - U'_i(n) + V_i(n-1)]
 \end{aligned}
 \tag{A11-C.14}$$

where

$$\begin{aligned}
 F_{\sin}(n) &\equiv Q_{L_{\sin}}(n-1) + Q'_{L_i}(n-1) + b_i(n)V_i(n-1) - V_{C_{\sin}}(n-1) \\
 F_{i+1}(n) &\equiv Q_{L_{i+1}}(n-1) - Q_{L_{\sin}}(n-1) - b_{i+1}(n)V_{i+1}(n-1) + V_{C_{\sin}}(n-1) \\
 H_{sf}(n) &= a_i(n) + b_i(n) + b_{\sin} \\
 H_{sb}(n) &= a_{i+1}(n) + b_{i+1}(n) + b_{\sin}
 \end{aligned}$$

Now, we consider the circuit representation at the bifurcation point of the vocal tract and the nasal tract. From Figure A11-C.3, the differential equations are given by

$$\begin{aligned}
 P_i(t) - P_{NC}(t) &= \frac{d}{dt} [L_i(t)U_{NC}(t)] + R_i(t)U_{NC}(t) \\
 P_{NC}(t) - P_{i+1}(t) &= \frac{d}{dt} [L_{i+1}(t)U_{i+1}] + R_{i+1}(t)U_{i+1}(t) \\
 P_{NC}(t) - P_{N1}(t) &= \frac{d}{dt} [L_{N1}(t)U_{N1}] + R_{N1}(t)U_{N1}(t) \\
 U_i(t) - U_{NC}(t) &= u_{i1}(t) + u_{i2}(t) + u_{i3}(t) \\
 &= G_i(t)P_i(t) + \frac{d}{dt} \{G_i(t)P_i(t)\} + u_{i3}(t) \\
 U_{NC}(t) &= U_{i+1}(t) + U_{N1}(t)
 \end{aligned}
 \tag{A11-C.15}$$

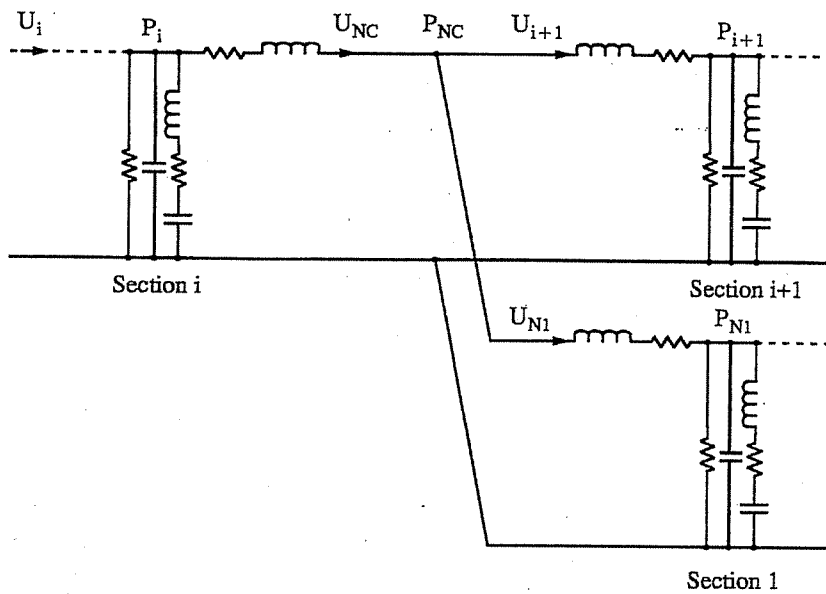


FIGURE A11-C.3 The circuit model of the bifurcation point.

After applying the discretization rules, previous derivation results, and some simple manipulations, one obtains the following difference equations

$$\begin{aligned}
 F_i(n) &= -b_{i-1}(n)U_{i-1}(n) + H_i(n)U_i(n) - b_i(n)U_{NC}(n) \\
 F_{NC}(n) &= -b_i(n)U_i(n) + H_{NC}(n)U_{NC}(n) + P_{NC}(n) \\
 F_{i+1}(n) &= -P_{NC}(n) + H_{i+1}(n)U_{i+1}(n) - b_{i+1}(n)U_{i+2}(n) \\
 P_i(n) &= b_i(n)[U_i(n) - U_{NC}(n) + V_i(n-1)] \\
 F_{N1}(n) &= -P_{NC}(n) + H_{N1}(n)U_{N1}(n) - b_{N1}(n)U_{N2}(n) \\
 U_{NC}(n) &= U_{i+1}(n) + U_{N1}(n)
 \end{aligned} \tag{A11-C.16}$$

where

$$\begin{aligned}
 F_i(n) &\equiv Q_{L_{i-1}L_i}(n-1) + b_{i-1}(n)V_{i-1}(n-1) - b_i(n)V_i(n-1) \\
 F_{NC}(n) &\equiv Q'_{L_i}(n-1) + b_i(n)V_i(n-1) \\
 F_{i+1}(n) &\equiv Q_{L_{i+1}}(n-1) - b_{i+1}(n)V_{i+1}(n-1) \\
 F_{N1}(n) &\equiv Q_{L_{N1}}(n-1) - b_{N1}(n)V_{N1}(n-1) \\
 H_{NC}(n) &= a_i(n) + b_i(n) \\
 H_{i+1}(n) &= a_{i+1}(n) + b_{i+1}(n) \\
 H_{N1}(n) &= a_{N1}(n) + b_{N1}(n) \\
 Q'_{L_i}(n-1) &= \frac{4}{T}L_i(n-1)U_{NC}(n-1) - Q'_{L_i}(n-2)
 \end{aligned}$$

Next, we consider the radiation part. Figure A11-C.4 shows the circuit model of the mouth section and radiation. The differential equations for this circuit are

$$\begin{aligned}
 P_i(t) - P_r(t) &= \frac{d}{dt}[L_r(t)U_r(t)] + R_r(t)U_r(t) \\
 U_r(t) &= \frac{P_r(t)}{R_r(t)} + \int_0^t \frac{P_r(\tau)}{L_r(\tau)} d\tau
 \end{aligned} \tag{A11-C.17}$$

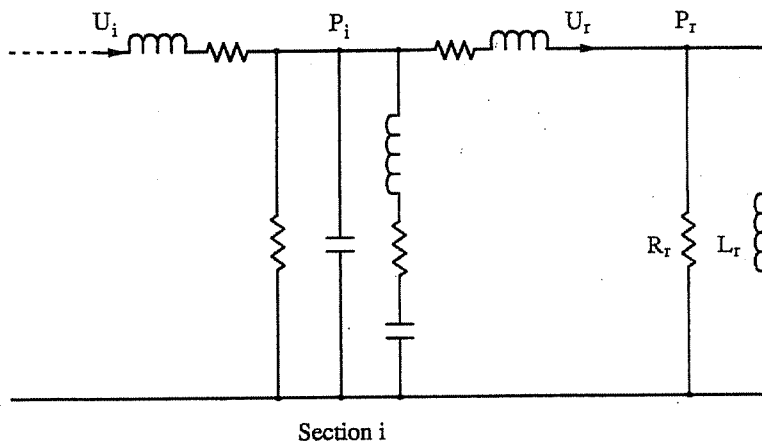


FIGURE A11-C.4 The circuit model of radiation.

The corresponding difference equations are

$$\begin{aligned}
 F_i(n) &= -b_{i-1}(n)U_{i-1}(n) + H_i(n)U_i(n) - b_i(n)U_r(n) \\
 F_r(n) &= -b_i(n)U_i(n) + H_r(n)U_r(n) \\
 P_i(n) &= b_i(n) [U_i(n) - U_r(n) + V_i(n-1)] \\
 P_r(n) &= b_r(n) [U_r(n) - V_{L_r}(n-1)]
 \end{aligned}
 \tag{A11-C.18}$$

where

$$\begin{aligned}
 F_i(n) &= Q_{L_{i-1}L_i}(n-1) + b_{i-1}(n)V_{i-1}(n-1) - b_i(n)V_i(n-1) \\
 F_r(n) &= Q'_{L_i}(n-1) + b_i(n)V_i(n-1) + b_r(n)V_{L_r}(n-1) \\
 H_r(n) &= a_i(n) + b_i(n) + b_r(n) \\
 b_r(n) &= \frac{1}{\frac{1}{R_r(n)} + \frac{T}{2} \frac{1}{L_r(n)}} \\
 Q'_{L_i}(n-1) &= \frac{4}{T} L_i(n-1)U_r(n-1) - Q_{L_i}(n-2) \\
 V_{L_r}(n-1) &= \frac{T}{L_r(n-1)} P_r(n-1) + V_{L_r}(n-2)
 \end{aligned}$$

Finally, we consider the excitation input. We have two types: one is without the glottal impedance and the subglottal system; one is with the glottal impedance and the subglottal system. We first derive the difference equations for no glottal impedance and no subglottal system. From Figure A11-C.5, it is easy to obtain the difference equations from previous derivation results.

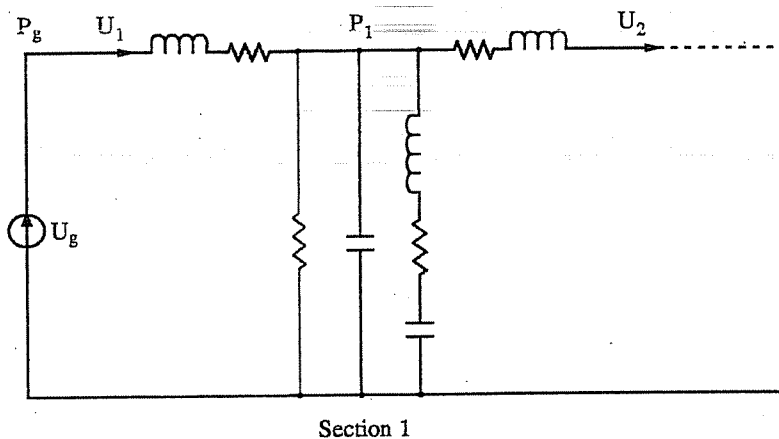


FIGURE A11-C.5 The circuit model of the first section of vocal tract with excitation.

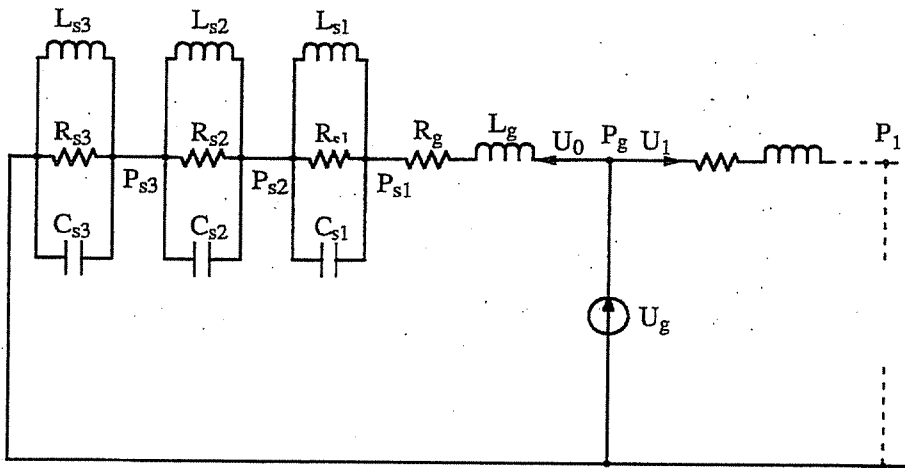


FIGURE A11-C.6 The circuit model of subglottal system and glottal impedance with excitation.

They are

$$\begin{aligned}
 F_1(n) &= -P_g(n) - b_1(n)U_2(n) \\
 U_1(n) &= U_g(n) \\
 P_1(n) &= b_1(n)[U_1(n) - U_2(n) + V_1(n-1)]
 \end{aligned}
 \tag{A11-C.19}$$

where

$$F_1(n) \equiv Q_{L_1}(n-1) + b_1(n)V_1(n-1) - [a_1(n) + b_1(n)]U_1(n)$$

For the case of excitation source with the glottal impedance and the subglottal system, we form the circuit model in Figure A11-C.6. The corresponding differential equations are given by

$$\begin{aligned}
 U_0(t) &= \frac{[P_{s1}(t) - P_{s2}(t)]}{R_{s1}} + \frac{1}{L_{s1}} \int_0^t [P_{s1}(\tau) - P_{s2}(\tau)] d\tau + C_{s1} \frac{d}{dt} [P_{s1}(t) - P_{s2}(t)] \\
 &= \frac{[P_{s2}(t) - P_{s3}(t)]}{R_{s2}} + \frac{1}{L_{s2}} \int_0^t [P_{s2}(\tau) - P_{s3}(\tau)] d\tau + C_{s2} \frac{d}{dt} [P_{s2}(t) - P_{s3}(t)] \\
 &= \frac{P_{s3}(t)}{R_{s3}} + \frac{1}{L_{s3}} \int_0^t P_{s3}(\tau) d\tau + C_{s3} \frac{dP_{s3}(t)}{dt}
 \end{aligned}
 \tag{A11-C.20}$$

$$P_g(t) - P_{s1}(t) = R_g(t)U_0(t) + \frac{d}{dt}[L_g(t)U_0(t)]$$

$$U_g(t) = U_0(t) + U_1(t)$$

Applying the discretization rules to above equations and solving  $P_{si}(n)$ ,  $i = 1, 2, 3$ , in order, then the difference equations can be written as

$$\begin{aligned}
 F_s(n) &= P_g(n) + X_{\text{sub}}(n)U_1(n) \\
 U_g(n) &= U_1(n) + U_0(n) \\
 F_1(n) &= -P_g(n) + H_1(n)U_1(n) - b_1(n)U_2(n)
 \end{aligned}
 \tag{A11-C.21}$$

where

$$\begin{aligned}
 F_s(n) &\equiv V_{s1}(n-1) + V_{s2}(n-1) + V_{s3}(n-1) + X_{\text{sub}}(n)U_g(n) - Q_{L_s}(n-1) \\
 F_1(n) &\equiv Q_{L_1}(n-1) - b_1(n)V_1(n-1) \\
 V_{si}(n-1) &= b_{si} \left[ C_{si}Q_{C_{si}}(n-1) - \frac{V_{L_{si}}(n-1)}{L_{si}} \right], \quad i = 1, 2, 3
 \end{aligned}$$

$$\begin{aligned}
b_{si}(n) &= \frac{1}{\frac{1}{R_{si}(n)} + \frac{T}{2} \frac{1}{L_{si}(n)} + \frac{2}{T} C_{si}}, \quad i = 1, 2, 3 \\
Q_{C_{si}}(n-1) &= \frac{4}{T} [P_{si}(n-1) - P_{si+1}(n-1)] - Q_{C_{si}}(n-2), \quad i = 1, 2 \\
Q_{C_{s3}}(n-1) &= \frac{4}{T} P_{s3}(n-1) - Q_{C_{s3}}(n-2) \\
V_{L_{si}}(n-1) &= T [P_{si}(n-1) - P_{si+1}(n-1)] + V_{L_{si}}(n-2), \quad i = 1, 2 \\
V_{L_{s3}}(n-1) &= T P_{s3}(n-1) + V_{L_{s3}}(n-2) \\
Q_{C_{si}}(n-1) &= \frac{4}{T} [P_{si}(n-1) - P_{si+1}(n-1)] - Q_{C_{si}}(n-2), \quad i = 1, 2 \\
Q_{C_{s3}}(n-1) &= \frac{4}{T} P_{s3}(n-1) - Q_{C_{s3}}(n-2) \\
P_{si}(n) &= \sum_{k=i}^1 b_{sk} U_0(n) + \sum_{k=i}^1 V_{sk}(n-1), \quad i = 1, 2, 3 \\
P_g(n) &= V_{s1}(n-1) + V_{s2}(n-1) + V_{s3}(n-1) + X_{sub}(n) U_0(n) - Q_{L_g}(n-1) \\
Q_{L_g}(n-1) &= \frac{4}{T} [L_g(n-1) U_0(n-1)] - Q_{L_g}(n-2) \\
X_{sub}(n) &= a_g(n) + b_{s1} + b_{s2} + b_{s3} \\
a_g(n) &= R_g(n) + \frac{2}{T} L_g(n)
\end{aligned}$$

We have derived the difference equations for different parts of the vocal system. Note that in order to set all recursive equations,  $Q(n-1)$  and  $V(n-1)$ , in motion, the initial rest conditions of the vocal system need to be assumed; that is,  $Q(0) = 0$  and  $V(0) = 0$  for all sections. To write out the matrix equations for the entire vocal system, we assume that

1. The vocal tract has a number of sections denoted as NING.
2. The nasal tract has a number of sections denoted as NTN.
3. The bifurcation point is located at the downstream of a vocal tract section denoted as NTS.
4. The excitation source is located at the input to first section of vocal tract.
5. One sinus is inserted at the downstream of a nasal-tract section denoted as NS.

For the vocal system that has no glottal impedance and no subglottal system but has one nasal sinus coupling, the three sets of simultaneous difference equations or the matrix equations for the pharyngeal, oral, and nasal tracts, respectively, are

$$\begin{aligned}
U_1(n) &= U_g(n) \\
k = 0: F_1(n) &= Q_{L_1}(n-1) - b_1(n) V_1(n-1) - H_1(n) U_1(n) \\
&= -P_g(n) - b_1(n) U_2(n) \\
k = 1: F_2(n) &= b_{\bar{1}}(n) V_1(n-1) + Q_{L_1 L_2}(n-1) - b_2(n) V_2(n-1) + b_1(n) U_1(n) \\
&= H_2(n) U_2(n) - b_2(n) U_3(n) \\
2 \leq k < \text{NTS}: F_{k+1}(n) &= b_{\bar{k}}(n) V_k(n-1) + Q_{L_k L_{k+1}}(n-1) - b_{k+1}(n) V_{k+1}(n-1) \\
&= -b_k(n) U_k(n) + H_{k+1}(n) U_{k+1}(n) - b_{k+1}(n) U_{k+2}(n) \\
&\quad \text{if } k+2 = \text{NTS} + 1 \Rightarrow U_{k+2}(n) = U_{\text{NC}}(n) \\
k = \text{NTS}: F_{\text{NC}}(n) &= Q'_{L_{\text{NTS}}}(n-1) + b_{\text{NTS}}(n) V_{\text{NTS}}(n-1) \\
&= -b_{\text{NTS}}(n) U_{\text{NTS}}(n) + H_{\text{NC}}(n) U_{\text{NC}}(n) + P_{\text{NC}}(n) \\
P_i(n) &= b_i(n) [U_i(n) - U_{i+1}(n) + V_i(n)], \quad \text{where } i = 1, \dots, \text{NTS}
\end{aligned}$$

if

$$i = \text{NTS},$$

then

$$U_{NTS+1}(n) = U_{NC}(n)$$

$$\begin{bmatrix} F_1 \\ F_2 \\ \vdots \\ F_{NTS} \\ F_{NC} \end{bmatrix} = \begin{bmatrix} -1 & -b_1 & 0 & \cdots & 0 & 0 & 0 \\ 0 & H_2 & -b_2 & \cdots & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & \cdots & H_{NTS} & -b_{NTS} & 0 \\ 0 & 0 & 0 & \cdots & -b_{NTS} & H_{NC} & 1 \end{bmatrix} \begin{bmatrix} P_g \\ U_2 \\ \vdots \\ U_{NC} \\ P_{NC} \end{bmatrix} \quad (A11-C.22)$$

$$\begin{aligned} F_{NTS+1}(n) &\equiv Q_{L_{NTS+1}}(n-1) - b_{NTS+1}(n)V_{NTS+1}(n-1) \\ &= -P_{NC}(n) + H_{NTS+1}(n)U_{NTS+1}(n) - b_{NTS+1}(n)U_{NTS+2}(n) \\ F_i(n) &\equiv Q_{L_{i-1}L_i}(n-1) + b_{i-1}(n)V_{i-1}(n-1) - b_i(n)V_i(n-1) \\ &= -b_{i-1}(n)U_{i-1}(n) + H_i(n)U_i(n) - b_i(n)U_{i+1}(n) \end{aligned}$$

where

$$i = NTS + 2, \dots, NING, \quad \text{if } i = NING$$

then

$$\begin{aligned} U_{NING+1}(n) &= U_r(n) \\ F_r(n) &\equiv b_{NING}(n)V_{NING}(n-1) + b_r(n)V_{L_r}(n-1) + Q'_{NING}(n-1) \\ &= -b_{NING}(n)U_{NING}(n) + H_r(n)U_r(n) \\ P_i(n) &= b_i(n)[U_i(n) - U_{i+1}(n) + V_i(n-1)], \quad \text{where } i = NTS + 1, \dots, NING - 1 \\ P_{NING}(n) &= b_{NING}(n)[U_{NING}(n) - U_r(n) + V_{NING}(n-1)] \\ P_r(n) &= b_r(n)[U_r(n) - V_{L_r}(n-1)] \end{aligned}$$

$$\begin{bmatrix} F_{NTS+1} \\ F_{NTS+2} \\ F_{NTS+3} \\ \vdots \\ F_{NING} \\ F_r \end{bmatrix} = \begin{bmatrix} -1 & H_{NTS+1} & -b_{NTS+1} & 0 & \cdots & 0 & 0 & 0 \\ 0 & -b_{NTS+1} & H_{NTS+2} & -b_{NTS+2} & \cdots & 0 & 0 & 0 \\ 0 & 0 & -b_{NTS+2} & H_{NTS+3} & \cdots & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & \cdots & -b_{NING-1} & H_{NING} & -b_{NING} \\ 0 & 0 & 0 & 0 & \cdots & 0 & -b_{NING} & H_r \end{bmatrix} \begin{bmatrix} P_{NC} \\ U_{NTS+1} \\ U_{NTS+2} \\ \vdots \\ U_{NING} \\ U_r \end{bmatrix} \quad (A11-C.23)$$

$$\begin{aligned} F_{N1}(n) &= Q_{L_{N1}}(n-1) - b_{N1}(n)V_{N1}(n-1) \\ &= -P_{NC}(n) + H_{N1}(n)U_{N1}(n) - b_{N1}(n)U_{N2}(n) \\ F_{Ni}(n) &= Q_{L_{Ni-1}L_{Ni}}(n-1) + b_{Ni-1}(n)V_{Ni-1}(n-1) - b_{Ni}(n)V_{Ni}(n-1) \\ &= -b_{Ni-1}(n)U_{Ni-1}(n) + H_{Ni}(n)U_{Ni}(n) - b_{Ni}(n)U_{Ni+1}(n) \\ i &= 2, \dots, N_s, N_s+2, \dots, NTN; \end{aligned}$$

if

$$i = N_s$$

then

$$\begin{aligned} U_{NNs+1}(n) &= U'_{NNs}(n); \quad \text{if } i = NTN, \text{ then } U_{NNNTN+1}(n) = U_{Nr}(n) \\ F_{sin}(n) &= Q_{L_{sin}}(n-1) + Q'_{L_{NNs}}(n-1) + b'_{NNs}(n)V_{NNs}(n-1) - V_{C_{sin}}(n-1) \\ &= -b_{NNs}(n)U_{NNs}(n) + H_{sf}(n)U'_{NNs}(n) - b_{sin}U_{NNs+1}(n) \end{aligned}$$



$$\begin{aligned}
 F_{N_{Ns+1}}(n) &= Q_{L_{N_{Ns+1}}}(n-1) - Q_{L_{sin}}(n-1) + V_{C_{sin}}(n-1) - b_{N_{Ns+1}}(n)V_{N_{Ns+1}}(n-1) \\
 &= -b_{sin}U'_{N_{Ns}}(n) + H_{sb}(n)U_{N_{Ns+1}}(n) - b_{N_{Ns+1}}(n)U_{N_{Ns+2}}(n) \\
 F_{N_r}(n) &= b_{N_{NTN}}(n)V_{N_{NTN}}(n-1) + b_{N_r}(n)V_{L_{N_r}}(n-1) + Q'_{L_{N_{NTN}}}(n-1) \\
 &= -b_{N_{NTN}}(n)U_{N_{NTN}}(n) + H_{N_r}(n)U_{N_r}(n) \\
 P_{N_i}(n) &= b_{N_i}(n)[U_{N_i}(n) - U_{N_{i+1}}(n) + V_{N_i}(n-1)],
 \end{aligned}$$

where

$$i = 1, 2, \dots, NTN; \text{ if } i = N_s, \text{ then } U_{N_{Ns+1}}(n) = U'_{N_{Ns}}(n)$$

if

$$i = NTN, \text{ then } U_{N_{NTN+1}}(n) = U_{N_r}(n)$$

$$\begin{aligned}
 P_{sin}(n) &= b_{sin}[U'_{N_{Ns}}(n) - U_{N_{Ns+1}}(n)] - Q_{L_{sin}}(n-1) + V_{C_{sin}}(n-1) \\
 P_{N_r}(n) &= b_{N_r}(n)[U_{N_r}(n) - V_{L_{N_r}}(n-1)]
 \end{aligned}$$

$$\begin{bmatrix} F_{N_1} \\ F_{N_2} \\ F_{N_3} \\ \vdots \\ F_{sin} \\ F_{N_{Ns+1}} \\ \vdots \\ F_{N_{NTN}} \\ F_{N_r} \end{bmatrix} = \begin{bmatrix} -1 & H_{N_1} & -b_{N_1} & \cdots & 0 & 0 & 0 & \cdots & 0 & 0 \\ 0 & -b_{N_1} & H_{N_2} & \cdots & 0 & 0 & 0 & \cdots & 0 & 0 \\ 0 & 0 & -b_{N_2} & \cdots & 0 & 0 & 0 & \cdots & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & \cdots & -b_{N_{Ns}} & H_{sf} & -b_{sin} & \cdots & 0 & 0 \\ 0 & 0 & 0 & \cdots & 0 & -b_{sin} & H_{sb} & \cdots & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & \cdots & 0 & 0 & 0 & \cdots & H_{N_{NTN}} & -b_{N_{NTN}} \\ 0 & 0 & 0 & \cdots & 0 & 0 & 0 & \cdots & -b_{N_{NTN}} & H_{N_r} \end{bmatrix} \begin{bmatrix} P_{NC} \\ U_{N_1} \\ U_{N_2} \\ \vdots \\ U'_{N_{Ns}} \\ U_{N_{Ns+1}} \\ \vdots \\ U_{N_{NTN}} \\ U_{N_r} \end{bmatrix}$$

(A11-C.24)

If there is no nasal sinus cavity, the matrix Equation (A11-C.24) reduces to the following.

$$\begin{aligned}
 F_{N_1}(n) &= Q_{L_{N_1}}(n-1) - b_{N_1}(n)V_{N_1}(n-1) \\
 &= -P_{NC}(n) + H_{N_1}(n)U_{N_1}(n) - b_{N_1}(n)U_{N_2}(n) \\
 F_{N_i}(n) &= Q_{L_{N_{i-1}L_{N_i}}}(n-1) + b_{N_{i-1}}(n)V_{N_{i-1}}(n-1) - b_{N_i}(n)V_{N_i}(n-1) \\
 &= -b_{N_{i-1}}(n)U_{N_{i-1}}(n) + H_{N_i}(n)U_{N_r}(n) - b_{N_i}(n)U_{N_{i+1}}(n)
 \end{aligned}$$

where

$$i = 2, \dots, NTN; \text{ if } i = NTN, \text{ then } U_{N_{NTN+1}}(n) = U_{N_r}(n)$$

$$\begin{aligned}
 F_{N_r}(n) &= b_{N_{NTN}}(n)V_{N_{NTN}}(n-1) + b_{N_r}(n)V_{L_{N_r}}(n-1) + Q'_{L_{N_{NTN}}}(n-1) \\
 &= -b_{N_{NTN}}(n)U_{N_{NTN}}(n) + H_{N_r}(n)U_{N_r}(n) \\
 P_{N_i}(n) &= b_{N_i}(n)[U_{N_i}(n) - U_{N_{i+1}}(n) + V_{N_i}(n-1)],
 \end{aligned}$$

where

$$i = 1, \dots, NTN; \text{ if } i = NTN, \text{ then } U_{N_{NTN+1}}(n) = U_{N_r}(n)$$

$$P_{N_r}(n) = b_{N_r}(n)[U_{N_r}(n) - V_{L_{N_r}}(n-1)]$$

$$\begin{bmatrix} F_{N_1} \\ F_{N_2} \\ F_{N_3} \\ \vdots \\ \vdots \\ F_{N_{NTN}} \\ F_{N_r} \end{bmatrix} = \begin{bmatrix} -1 & H_{N_1} & -b_{N_1} & 0 & \cdots & 0 & 0 & 0 \\ 0 & -b_{N_1} & H_{N_2} & -b_{N_2} & \cdots & 0 & 0 & 0 \\ 0 & 0 & -b_{N_2} & H_{N_3} & \cdots & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & \cdots & -b_{N_{NTN-1}} & H_{N_{NTN}} & -b_{N_{NTN}} \\ 0 & 0 & 0 & 0 & \cdots & 0 & -b_{N_{NTN}} & H_{N_r} \end{bmatrix} \begin{bmatrix} P_{NC} \\ U_{N_1} \\ U_{N_2} \\ \vdots \\ \vdots \\ U_{N_{NTN}} \\ U_{N_r} \end{bmatrix}$$

(A11-C.25)

If the glottal impedance and the subglottal system are included, the matrix Equation (A11-C.22) extends to the following.

$$\begin{bmatrix} F_s \\ F_1 \\ F_2 \\ \cdot \\ \cdot \\ \cdot \\ F_{NTS} \\ F_{NC} \end{bmatrix} = \begin{bmatrix} 1 & X_{sub} & 0 & \cdots & 0 & 0 & 0 \\ -1 & H_1 & -b_1 & \cdots & 0 & 0 & 0 \\ 0 & -b_1 & H_2 & \cdots & 0 & 0 & 0 \\ \cdot & \cdot & \cdot & \cdot & \cdot & \cdot & \cdot \\ \cdot & \cdot & \cdot & \cdot & \cdot & \cdot & \cdot \\ \cdot & \cdot & \cdot & \cdot & \cdot & \cdot & \cdot \\ 0 & 0 & 0 & \cdots & H_{NTS} & -b_{NTS} & 0 \\ 0 & 0 & 0 & \cdots & -b_{NTS} & H_{NC} & 1 \end{bmatrix} \begin{bmatrix} P_g \\ U_1 \\ U_2 \\ \cdot \\ \cdot \\ \cdot \\ U_{NC} \\ P_{NC} \end{bmatrix} \quad (\text{A11-C.26})$$

From the derived discrete-time acoustic matrix equations, we can form four different vocal system model structures, which are:

1. (A11-C.22), (A11-C.23), and (A11-C.24) for the vocal system model with nasal sinus but no glottal impedance and no subglottal system.
2. (A11-C.22), (A11-C.23), and (A11-C.25) for the vocal system model with no nasal sinus and no glottal impedance and no subglottal system.
3. (A11-C.26), (A11-C.23), and (A11-C.24) for the vocal system model with nasal sinus and with the glottal impedance and the subglottal system.
4. (A11-C.26), (A11-C.23), and (A11-C.25) for the vocal system model with no nasal sinus but with the glottal impedance and the subglottal system.

There are three matrix equations for each structure. Each matrix equation can be written as  $y = A \cdot x$ , where  $A$  is a non-square band diagonal sparse matrix of coefficients,  $y$  is a column vector of force constants, and  $x$  is the unknown column vector. The elements in the sparse coefficient matrix and the force constants inside the vocal system are defined in the previous paragraphs. In addition to the three matrix equations, we need the boundary condition at the nasal coupling point,  $U_{NC}(n) = U_{i+1}(n) + U_{Ni}(n)$ . For Cases 3 and 4, we need one more boundary condition at the glottis,  $U_g(n) = U_1(n) + U_0(n)$ . It may also be noted that the three matrix equations for each case are coupled by the term  $P_{NC}(n)$  and boundary condition(s). As Maeda (1982a) pointed out, if we eliminate  $P_{NC}(n)$  and  $U_{NC}(n)$  analytically, the formulation results in an unstable system. Therefore,  $P_{NC}(n)$  and  $U_{NC}(n)$  are included in the unknown column vectors in order for stable solutions.

There are many methods that can be used to solve these sparse linear system equations. Since the coefficient matrices are band diagonal sparse matrices, an efficient elimination procedure followed by a substitution procedure can be used. Once the values of  $U_i(n)$  and  $U_{Ni}(n)$  for all  $i$ ,  $P_{NC}(n)$ , and  $U_{NC}(n)$  are solved, the pressure  $P_i(n)$  and  $P_{Ni}(n)$  for all  $i$ , and the volume velocity  $u_{i3}(n)$  and  $u_{Ni3}(n)$  for all  $i$  can be computed. Then, the force constants and coefficient matrices are updated for the next recursion.

## COMMENTS ON USAGE OF SIMULATED ANNEALING PROCEDURE

In Appendix 11, we covered the details of speech inverse filtering using the simulated annealing optimization algorithm. The annealing parameters control the performance of the optimization process. The various acoustic characteristics of the speech signal and target-frame selections affect the optimization process. The default values of the annealing parameters are not the best values for all cases. Although we do not know the combinations of the annealing parameters that perform well for the optimization process for different target frames, the following guidelines provide some rules for adjusting the appropriate annealing parameters.

- Step 1:** Set the desired nasalization extent and set the number of dimensions of the articulatory vector at the appropriate dimensions, for example,  $M = 8$  for front vowels;  $M = 9$  for nasalized front vowels;  $M = 11$  for middle, back vowels, and semivowels; and  $M = 12$  for nasalized vowels (see the descriptions in Section A11.3.3.1). Start the optimization process with the default initial articulatory vector and the default annealing parameters. If the error distance is less than 1% after the process stops, go to Step 5. If not, go to Step 2.
- Step 2:** Check if the current vocal tract shape (or cross-sectional area) is reasonable. If the shape is not reasonable go to Step 3. Otherwise, record the error distance as  $\epsilon_p$  and the current final temperature as  $T_p$ . Then set the initial temperature  $T = \text{floor}[T_p]$ . Start the optimization process again. If the new error distance is less than  $\epsilon_p$ , then this step is repeated until the error criterion is met. If not go to Step 4.
- Step 3:** Examine the vocal tract shape (articulatory vector) for the initial settings and all subsequent values. Several adjustments of the annealing parameters can be used. The following order of adjustments are recommended: raise the initial temperature, increase the value of the reduction factor, increase the total number of evaluations, and change the other annealing parameters as desired. Then begin the process and apply Step 2.
- Step 4:** Examine the vocal tract shape as described in Step 3. Increase the number of dimensions of the articulatory vector from  $M = 8$  to  $M = 11$  and start the process. Apply Step 2.
- Step 5:** Check the vocal tract shape with x-ray tracings or schematic vocal tract profiles as published in the literature or compare with vocal tract profiles as shown in Appendix 5. If the vocal tract outline is similar to those published, then the optimization process is done. If not, this may mean that the "ventriloquist effect" has occurred. One can adjust the settings of the nine articulatory sliders in the Articulatory Position Settings popup window so that the initial configuration becomes closer to the true outline. Then go to Step 1 to start the optimization process again.
- Step 6:** If the above steps fail to reduce the error criterion to a satisfactory value, then return to the target-frame selection phase and reselect the current target frame. Start the optimization process from the Step 1.

The above guidelines are illustrated in the following examples for the sentence, "We were away a year ago," spoken by two male subjects, A and B. Figures A11-D.1 and A11-D.2 show the optimized target frames for subjects A and B, respectively. See Appendix A11-A for the vowel results.

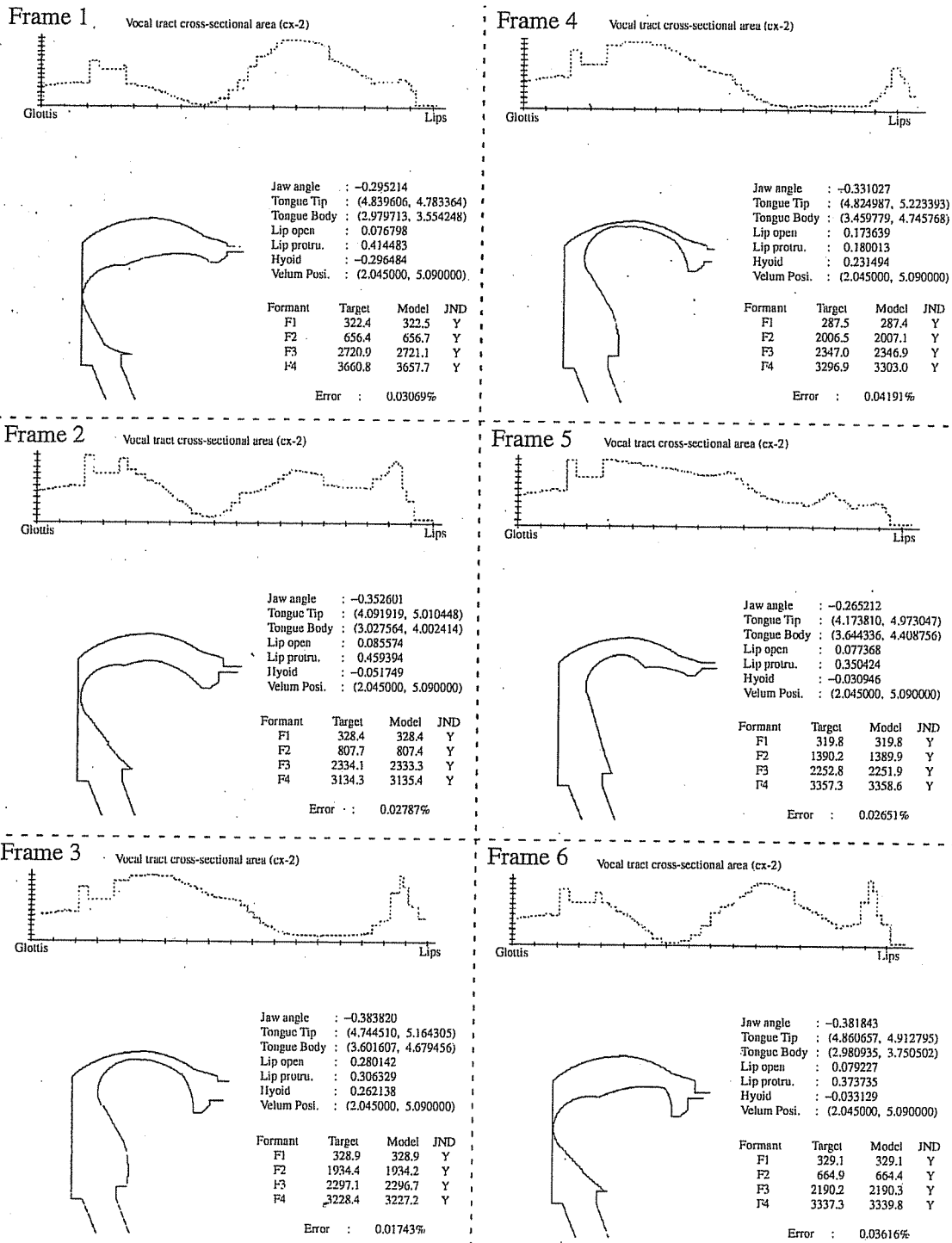


FIGURE A11-D.1 The optimized target frames of the sentence, "We were away a year ago," spoken by male subject A.

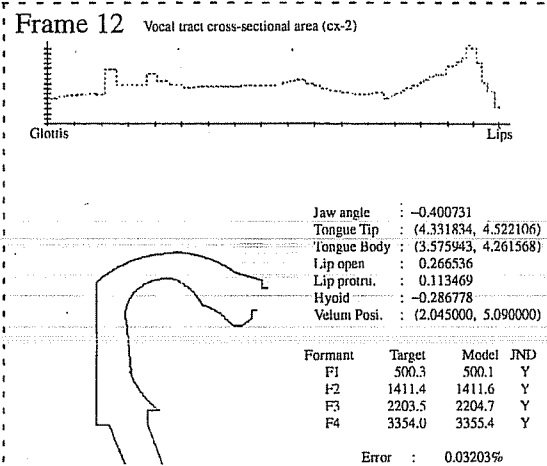
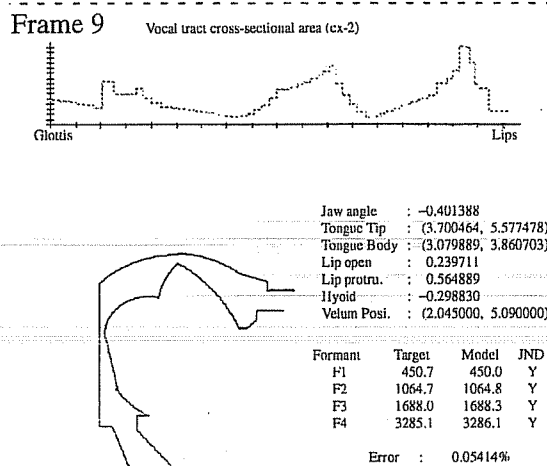
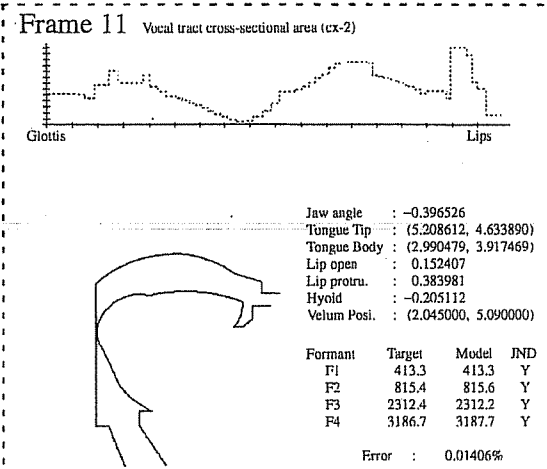
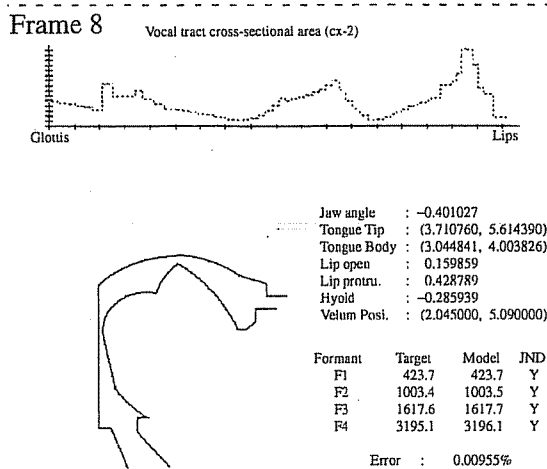
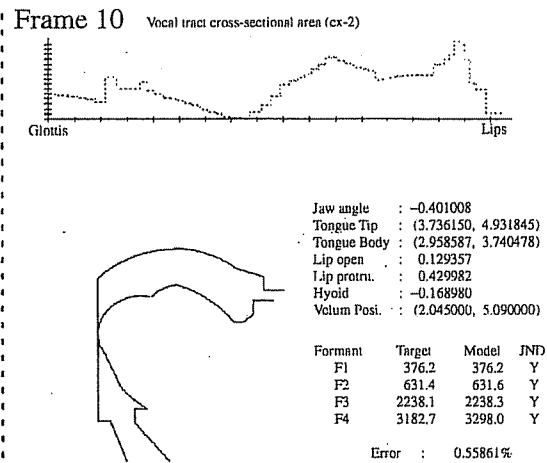
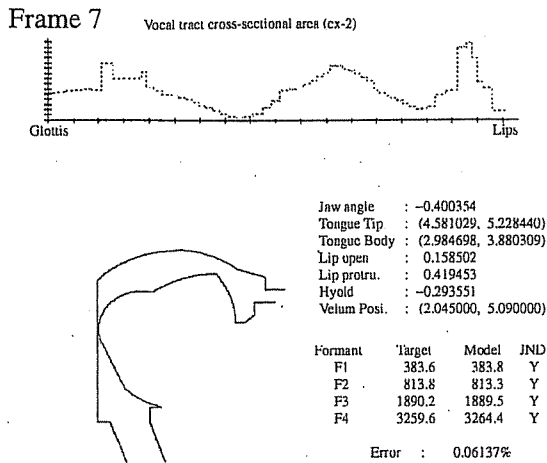


FIGURE A11-D.1 Continued.

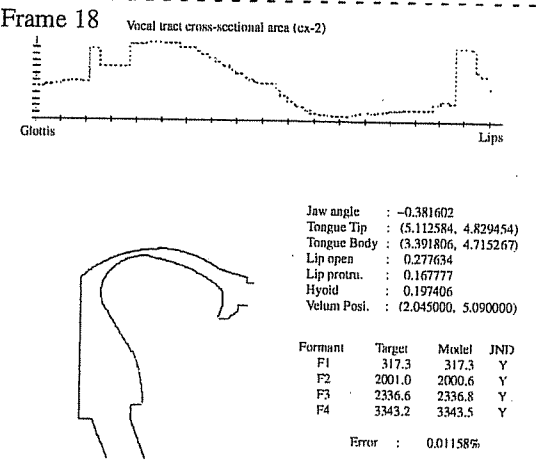
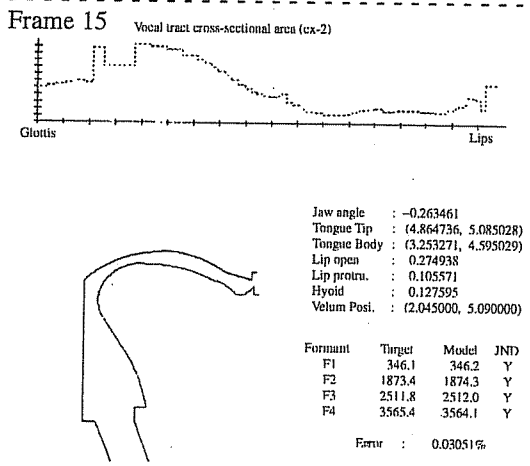
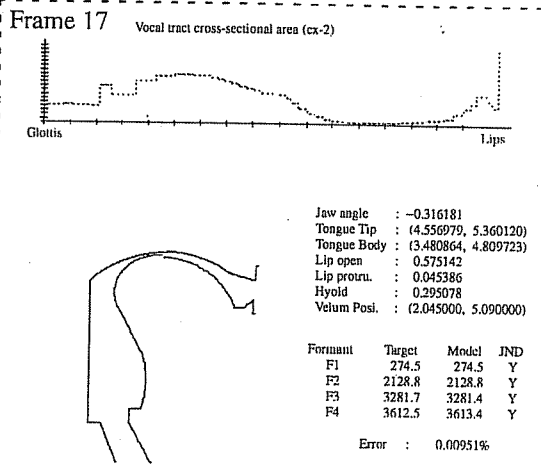
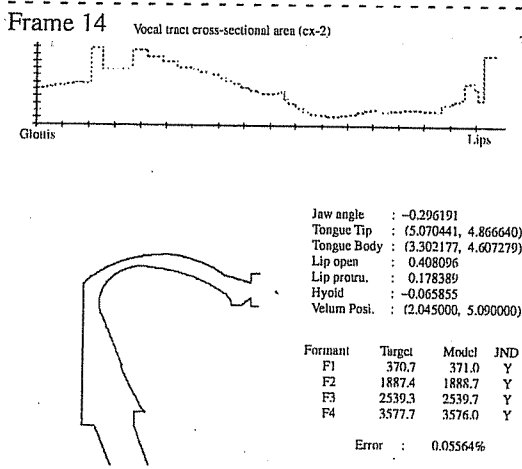
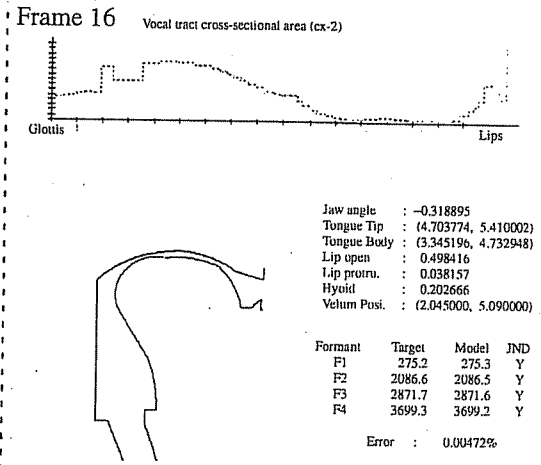
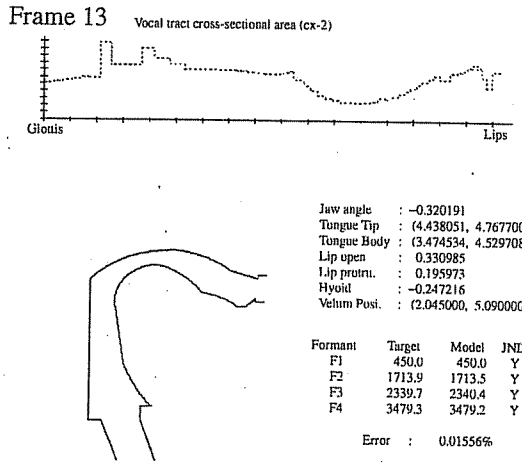


FIGURE A11-D.1 Continued.

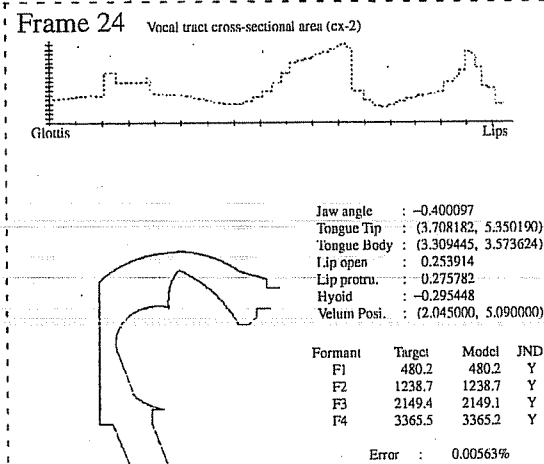
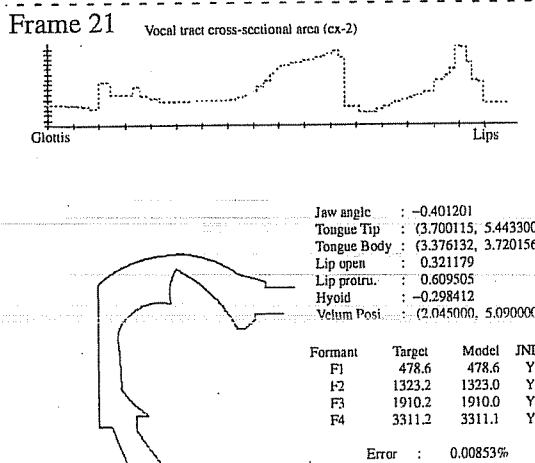
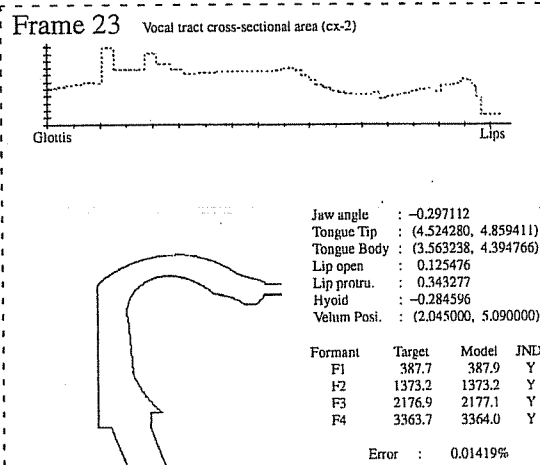
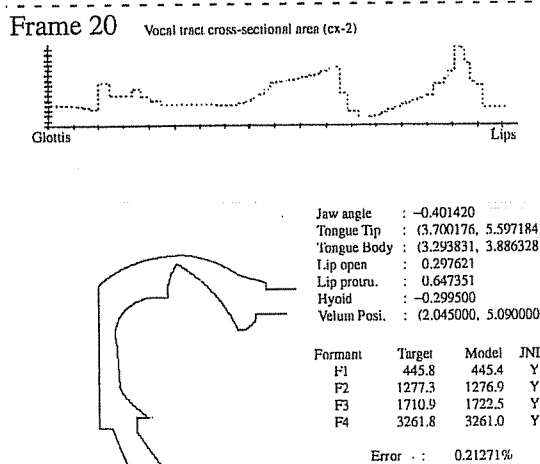
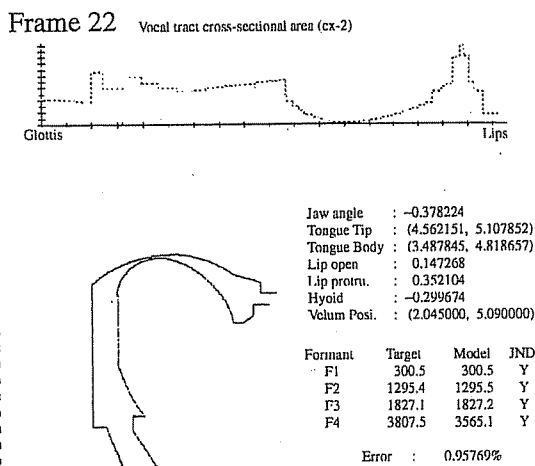
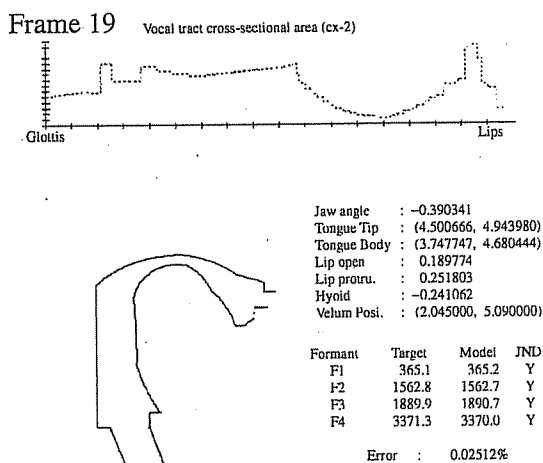


FIGURE A11-D.1 Continued.

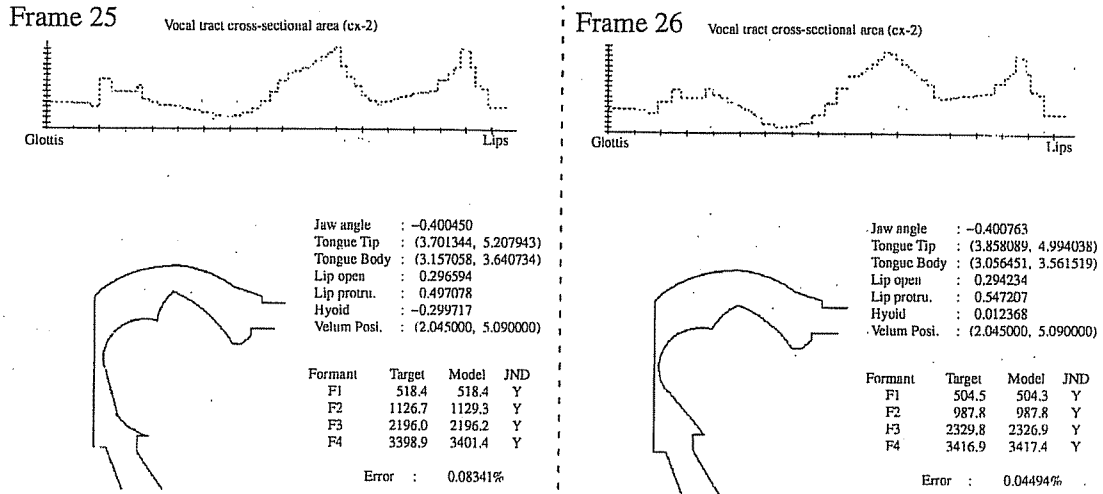


FIGURE A11-D.1 Continued.

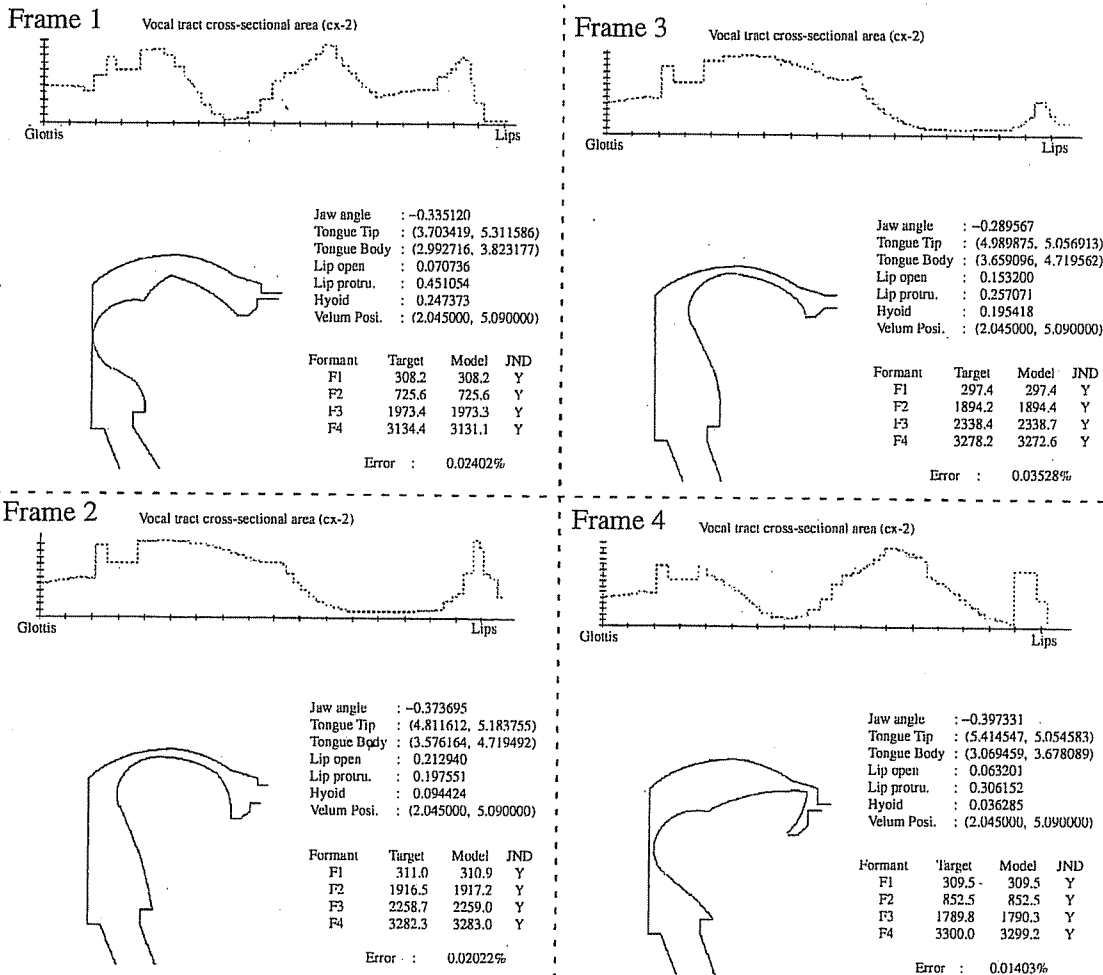


FIGURE A11-D.2 The optimized target frames of the sentence, "We were away a year ago," spoken by male subject B.



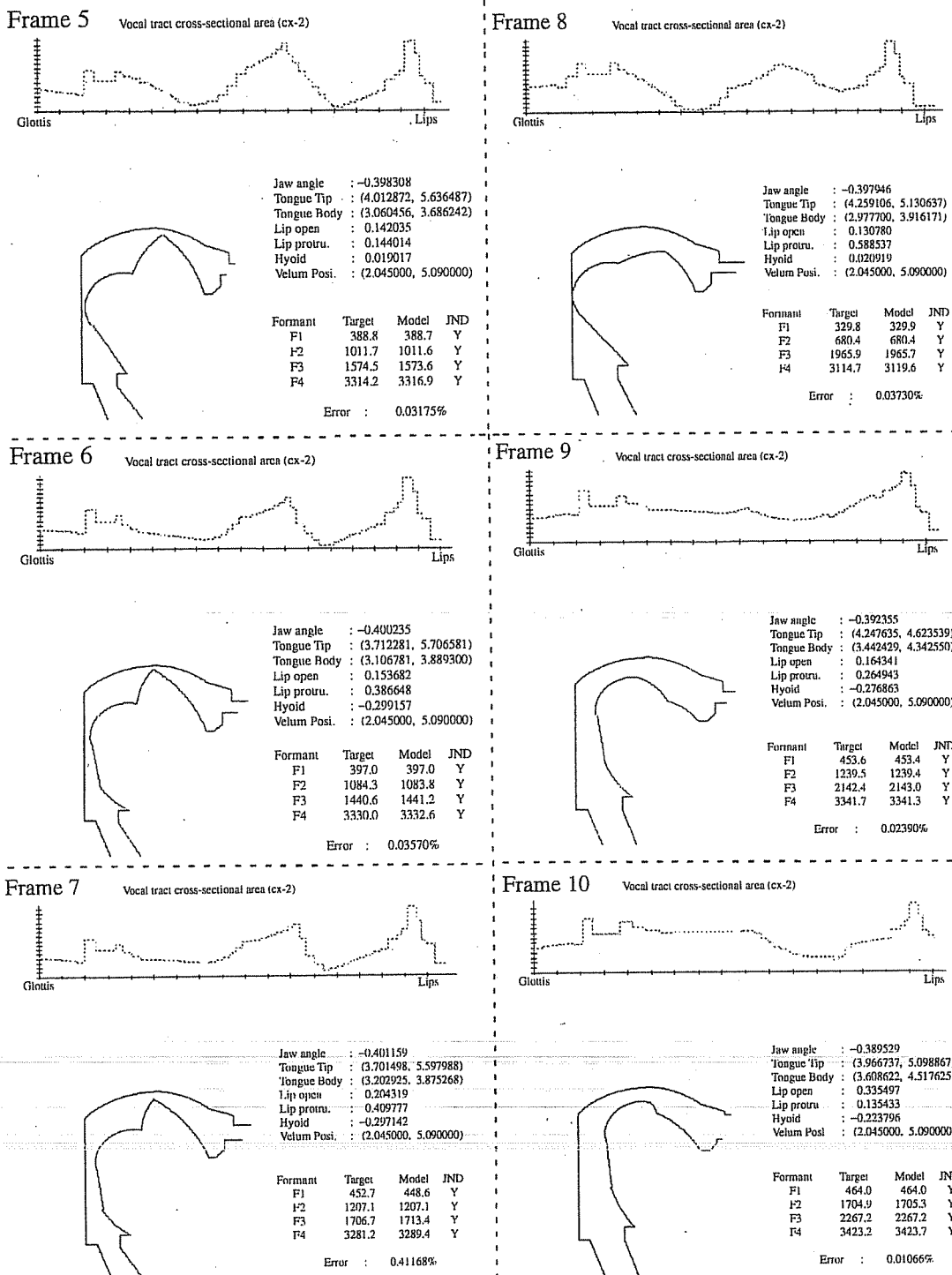


FIGURE A11-D.2 Continued.

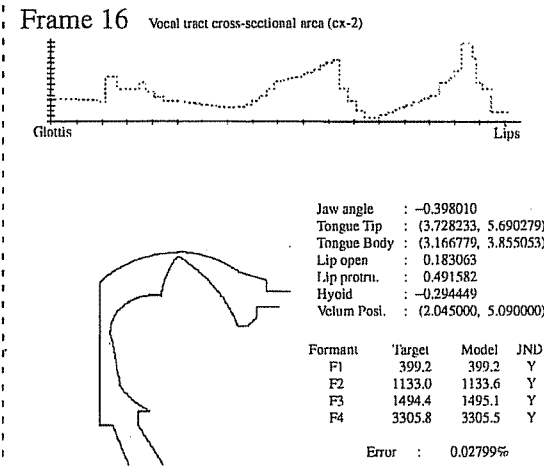
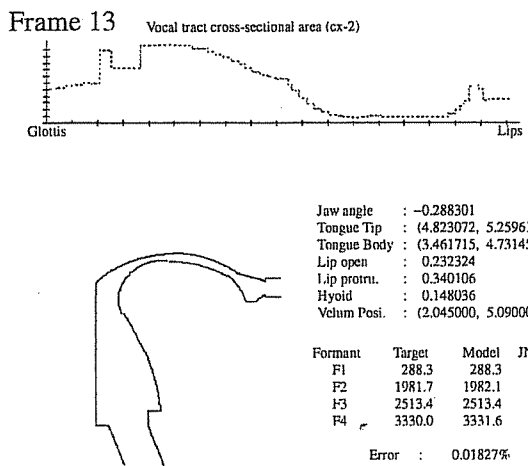
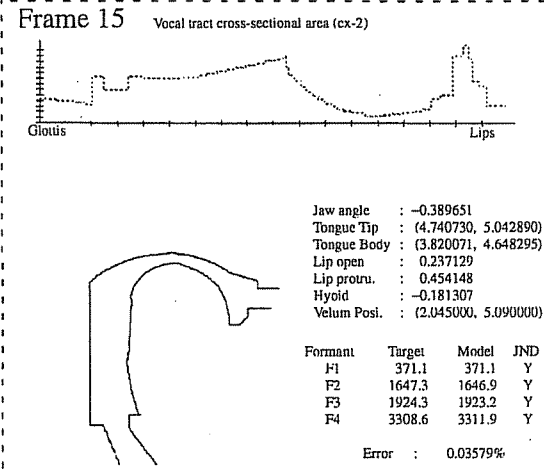
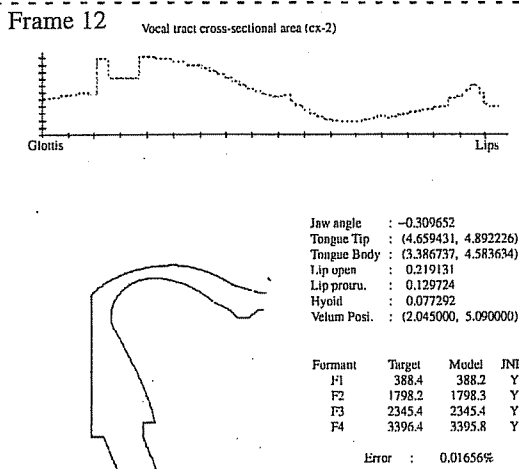
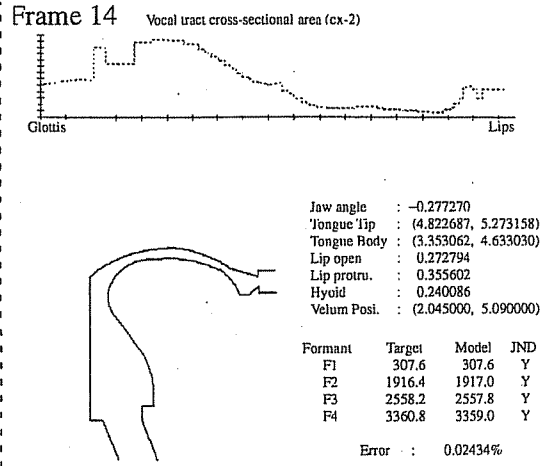
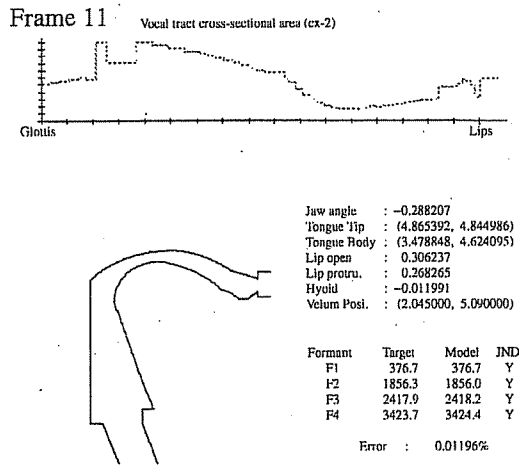


FIGURE A11-D.2 Continued.

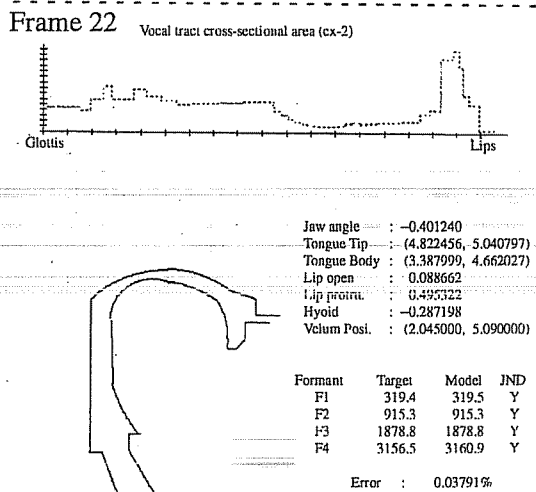
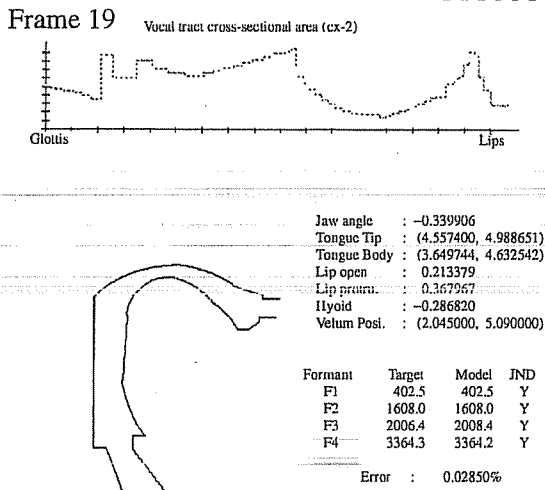
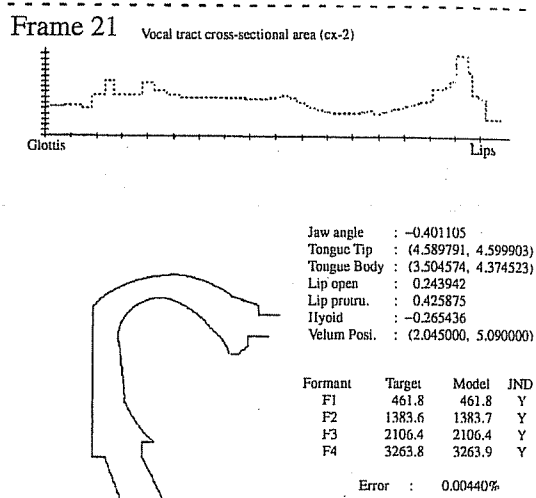
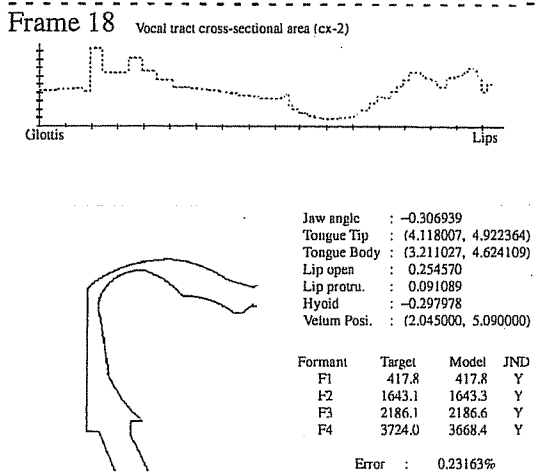
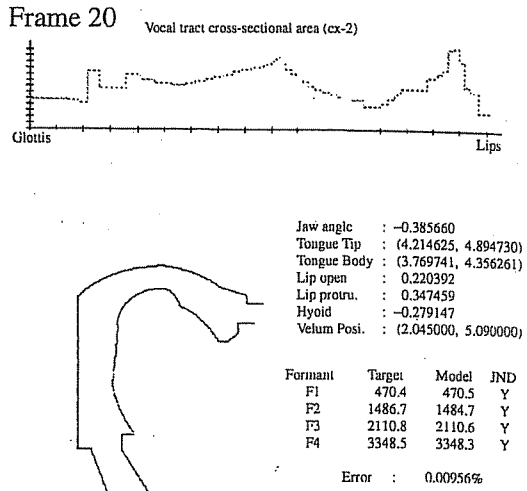
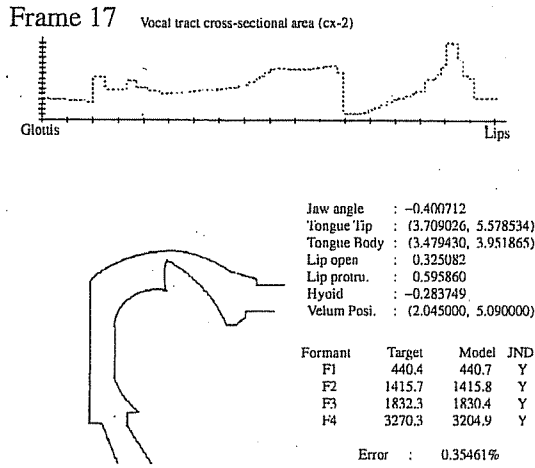


FIGURE A11-D.2 Continued.